

Chapter 12 :- Introduction to the Laplace Transform

The Laplace transform is given by the integral

$$L\{f(t)\} = F(s) = \int_{-\infty}^{\infty} f(t) e^{-st} dt$$

where the variable $s = \sigma + j\omega$

$f(t) \xleftrightarrow{L} F(s)$ Laplace transform pairs

$$L\{\delta(t)\} \leftrightarrow 1$$

please check Table 12.1 for other pairs and Table 12.2 for the properties. Please also see Table 12.3 for useful transform pairs.

Why Laplace :-

- Consider the steady-state and the transient response of circuits containing more than a single node.
- Transient response of a circuit containing complicated single source
- To define the transfer function
- Relate time domain to s -domain and frequency domain

Inverse Laplace transform

Let $F(s)$ be Laplace transform of some function $f(t)$
we want to find $f(t)$ without using the inversion
formula - we want to find $f(t)$ using the Laplace transform
known table and properties.

- put $F(s)$ in a form or a sum of forms that we know
it is in the Laplace transform table.

$F(s)$ in general is a ratio of two polynomials

$F(s)$ is a rational function

$$F(s) = \frac{N(s)}{D(s)} = \frac{a_n s^n + a_{n-1} s^{n-1} + \dots + a_0}{b_m s^m + b_{m-1} s^{m-1} + \dots + b_0}$$

If $m > n$, the ratio $\frac{N(s)}{D(s)}$ is called proper rational function

If $m \leq n$, the ratio $\frac{N(s)}{D(s)}$ is called improper rational function

Only a proper rational function can be expanded as a sum of
partial fraction.

Partial fraction expansion Non repeated Real Roots of DCs (Distinct)

$$F(s) = \frac{96(s+5)(s+12)}{s(s+8)(s+6)} \equiv \frac{k_1}{s} + \frac{k_2}{s+8} + \frac{k_3}{s+6}$$

To find the value of k_1 , we multiply both sides by s and then evaluate both sides at $s=0$

$$s \frac{96(s+5)(s+12)}{s(s+8)(s+6)} \Big|_{s=0} = s \frac{k_1}{s} \Big|_{s=0} + \frac{k_2 s}{s+8} \Big|_{s=0} + \frac{k_3 s}{s+6} \Big|_{s=0}$$

$$\frac{96(s+5)(s+12)}{(s+8)(s+6)} \Big|_{s=0} = k_1 + \frac{k_2 s}{s+8} \Big|_{s=0} + \frac{k_3 s}{s+6} \Big|_{s=0}$$

$$\frac{96(5)(12)}{(8)(6)} = k_1 + 0 + 0$$

$$\boxed{120 = k_1}$$

To find the value of k_2 , we multiply both sides by $(s+8)$ and then evaluate both sides at $s=-8$

$$\frac{96(s+5)(s+12)}{s(s+6)} \Big|_{s=-8} = \frac{k(s+8)}{s} \Big|_{s=-8} + k_2 + \frac{k_3(s+8)}{(s+6)} \Big|_{s=-8}$$

$$\frac{96(-3)(4)}{-8(-2)} = 0 + k_2 + 0$$

$$\boxed{72 = k_2}$$

then K_3

$$\frac{96(s+5)(s+12)}{s(s+8)} \Big|_{s=-6} = \frac{K_1(s+6)}{s} \Big|_{s=-6} + \frac{K_2(s+6)}{s+8} \Big|_{s=-6} + K_3$$

$$\frac{96(-1)(6)}{-6(2)} = 0 + 0 + K_3$$

$$48 = K_3$$

So,

$$\frac{96(s+5)(s+12)}{s(s+8)(s+6)} = \frac{120}{s} + \frac{48}{s+6} - \frac{72}{s+8}$$

$$\therefore \mathcal{L}^{-1} \left[\frac{96(s+5)(s+12)}{s(s+8)(s+6)} \right] = \underbrace{(120)}_{\text{steady}} + \underbrace{(48e^{-6t} - 72e^{-8t})}_{\text{transient}} u(t)$$

Partial Fraction Expansion: Repeated Real Roots of DCs

$$\frac{100(s+25)}{s(s+5)^3} = \frac{K_1}{s} + \frac{K_2}{(s+5)^3} + \frac{K_3}{(s+5)^2} + \frac{K_4}{(s+5)}$$

we find K_1 as previously described

$$\frac{100(s+25)}{(s+5)^3} \Big|_{s=0} = K_1 + \frac{K_2 s}{(s+5)^3} \Big|_{s=0} + \frac{K_3 s}{(s+5)^2} \Big|_{s=0} + \frac{K_4 s}{(s+5)} \Big|_{s=0}$$

$$\frac{100(25)}{5} = 20$$

To find k_2 , we multiply both sides by $(s+5)^3$ and then evaluate both sides at -5

$$\frac{100(s+25)}{s} = \frac{k_1(s+5)^3}{s} \Big|_{s=-5} + k_2 + k_3(s+5) \Big|_{s=-5} + k_4(s+5)^2 \Big|_{s=-5}$$

$$\frac{100(20)}{-5} = 0 + k_2 + 0 + 0$$

$$\boxed{-400 = k_2}$$

To find k_3 , we first multiply both sides by $(s+5)^3$. Next we differentiate both sides once with respect to s and then evaluate at $s = -5$

$$\frac{d}{ds} \left[\frac{100(s+25)}{s} \right] \Big|_{s=-5} = \frac{d}{ds} \left[\frac{k_1(s+5)^3}{s} \right] \Big|_{s=-5} + \frac{d}{ds} [k_2] + \frac{d}{ds} [k_3(s+5)] \Big|_{s=-5} + \frac{d}{ds} [k_4(s+5)^2] \Big|_{s=-5}$$

$$100 \left[\frac{s - (s+25)}{s^2} \right] \Big|_{s=-5} = k_3$$

$$\boxed{-100 = k_3}$$

To find k_4 , we first multiply both sides by $(s+5)^3$. Next we differentiate both sides with respect to s twice and then evaluate both sides at $s = -5$.

After simplifying the first derivative, the second derivative becomes

$$100 \frac{d}{ds} \left[\frac{-25}{s^2} \right] \Big|_{s=-5} = k_1 \frac{d}{ds} \left[\frac{(s+5)^2(2s-5)}{s^2} \right] \Big|_{s=-5} + 0 + \frac{d}{ds} [k_3] + \frac{d}{ds} [2k_4(s+5)] \Big|_{s=-5}$$

$$100 \left[\frac{25(2s)}{s^4} \right] \Big|_{s=-5} = -2K_4$$

$$\frac{(100)(-250)}{625} = 2K_4$$

$$-40 = 2K_4$$

$$\boxed{-20 = K_4}$$

So

$$\frac{100(s+25)}{s(s+5)^3} = \frac{20}{s} - \frac{400}{(s+5)^3} - \frac{100}{(s+5)^2} - \frac{20}{s+5}$$

$$\therefore \mathcal{L}^{-1} \left[\frac{100(s+25)}{s(s+5)^3} \right] = [20 - 200t^2 e^{-5t} - 100t e^{-5t} - 20e^{-5t}] u(t)$$

3 - Partial Fraction Expansion & Nonrepeated Complex Roots of $D(s)$

$$F(s) = \frac{100(s+3)}{(s+6)(s^2+6s+25)}$$

$F(s)$ is a proper rational function. Next we must find the roots of the quadratic term: $s^2+6s+25$

$$s = \frac{-b \pm \sqrt{b^2 - 4ac}}{2a}$$

$$s^2 + 6s + 25 = (s+3-j4)(s+3+j4)$$

$$\frac{100(s+3)}{(s+6)(s^2+6s+25)} = \frac{K_1}{s+6} + \frac{K_2}{s+3-j4} + \frac{K_3}{s+3+j4}$$

To find K_1 , K_2 and K_3 we use the same process as before:

$$\frac{100(s+3)}{s^2+6s+25} \Big|_{s=-6} = K_1 + \frac{K_2(s+6)}{s+3-j4} \Big|_{s=-6} + \frac{K_3(s+6)}{s+3+j4} \Big|_{s=-6}$$

$$\frac{100(-3)}{25} = K_1$$

$$\boxed{-12 = K_1}$$

To find K_2 , we multiply both sides by $s+3-j4$, then evaluate at $s = -3+j4$

$$\frac{100(s+3)}{(s+6)(s+3+j4)} \Big|_{s=-3+j4} = K_2$$

$$\frac{100(-3+j4+3)}{(-3+j4+6)(-3+j4+3+j4)} = K_2$$

$$\frac{100(j4)}{(3+j4)(+j8)} = K_2$$

$$\boxed{\begin{aligned} 6-8j &= K_2 \\ 10e^{-j53.13} &= K_2 \end{aligned}}$$

To find K_3 , we multiply both sides by $s+3+j4$ and evaluate at $s = -3-j4$

$$\frac{100(s+3)}{(s+6)(s+3-j4)} \Big|_{s=-3-j4} = K_3$$

$$\frac{100(-j4)}{(3-j4)(-j8)} = K_3$$

$$\boxed{\begin{aligned} 6+j8 &= K_3 \\ 10e^{+j53.13} &= K_3 = K_2^* \end{aligned}}$$

$$\frac{100(s+3)}{(s+6)(s^2+6s+25)} = \frac{-12}{s+6} + \frac{10 \sqrt{-53.13}}{s+3-j4} + \frac{10 \sqrt{53.13}}{s+3+j4}$$

$$\mathcal{L}^{-1} \left[\frac{100(s+3)}{(s+6)(s^2+6s+25)} \right] = \left[-12e^{-6t} + 10e^{-j53.13-(3-j4)t} + 10e^{j53.13-(3+j4)t} \right] u(t)$$

$$10e^{-j53.13} e^{-(3-j4)t} + 10e^{j53.13} e^{-(3+j4)t}$$

$$= 10e^{-3t} \left(e^{j(4t-53.13)} + e^{-j(4t-53.13)} \right)$$

$$= 20e^{-3t} \cos(4t-53.13)$$

$$\therefore \mathcal{L}^{-1} \left[\frac{k}{s+\alpha-j\beta} + \frac{k^*}{s+\alpha+j\beta} \right] = 2|k| e^{-\alpha t} \cos(\beta t + \theta)$$

4- Partial Fraction Expansion: Repeated Complex Roots of $D(s)$

$$F(s) = \frac{768}{(s^2+6s+25)^2} = \frac{768}{(s+3-j4)^2 (s+3+j4)^2} \equiv$$

$$\equiv \frac{k_1}{(s+3-j4)^2} + \frac{k_2}{(s+3-j4)} + \frac{k_3}{(s+3+j4)^2} + \frac{k_4}{(s+3+j4)}$$

$$k_3 = k_1^*, \quad k_4 = k_2^*$$

to find k_1 , multiply both sides by $(s+3-j4)^2$ and then evaluate at $s = -3+j4$

$$\left. \frac{768}{(s+3+j4)^2} \right|_{s=-3+j4} = k_1$$

$$\frac{768}{(j8)^2} = k_1$$

$$\boxed{-12 = k_1} \Rightarrow k_3 = k_1^* = -12$$

To find k_2

$$\frac{d}{ds} \left[\frac{768}{(s+3+j4)^2} \right]_{s=-3+j4} = k_2$$

$$= \frac{-2(768)(s+3+j4)}{(s+3+j4)^3} = k_2$$

$$- \frac{2(768)}{(j8)^3} = k_2$$

$$-j3 = k_2$$

$$\boxed{3 \angle -90 = k_2} \Rightarrow k_4 = k_2^* = 3 \angle 90$$

$$F(s) = \left[\frac{-12}{(s+3-j4)^2} - \frac{12}{(s+3+j4)^2} \right] + \left(\frac{3 \angle -90}{s+3-j4} + \frac{3 \angle 90}{s+3+j4} \right)$$

$$f(t) = \left[-12t e^{-(3+j4)t} - 12t e^{-(3-j4)t} + 3 e^{-j90} e^{-(3+j4)t} + 3 e^{j90} e^{-(3-j4)t} \right] u(t)$$

$$= \left[-12t e^{-3t} e^{-j4t} - 12t e^{-3t} e^{+j4t} + 3 e^{-j90} e^{-3t} e^{-j4t} + 3 e^{j90} e^{-3t} e^{+j4t} \right] u(t)$$

$$= \left[-12t e^{-3t} (e^{j4t} + e^{-j4t}) + 3 e^{-3t} (e^{j(4t+90)} + e^{-j(4t+90)}) \right] u(t)$$

$$= \left[-24t e^{-3t} \cos(4t) + 6 e^{-3t} \cos(4t+90) \right] u(t)$$

$$\mathcal{L}^{-1}\left(\frac{k}{(s+a)^r}\right) = \frac{k t^{r-1} e^{-at}}{(r-1)!} u(t)$$

$$\mathcal{L}^{-1}\left(\frac{k}{(s+\alpha-j\beta)^r} + \frac{k^*}{(s+\alpha+j\beta)^r}\right) = \left[\frac{2|k| t^{r-1}}{(r-1)!} e^{-\alpha t} \cos(\beta t + \theta)\right] u(t)$$

5- Partial Fraction Expansion & Improper Rational Function

$$F(s) = \frac{s^4 + 13s^3 + 66s^2 + 200s + 300}{s^2 + 9s + 20}$$

$$\begin{array}{r} s^2 + 9s + 20 \overline{) s^4 + 13s^3 + 66s^2 + 200s + 300} \\ \underline{s^4 + 9s^3 + 20s^2} \\ 0 - 4s^3 + 46s^2 + 200s + 300 \\ \underline{4s^3 + 36s^2 + 80s} \\ 0 + 10s^2 + 120s + 300 \\ \underline{10s^2 + 90s + 200} \\ 0 + 30s + 100 \end{array}$$

$$\therefore F(s) = s^2 + 4s + 10 + \frac{30s + 10}{s^2 + 9s + 20}$$

where $\frac{30s + 10}{s^2 + 9s + 20}$ is a proper rational function

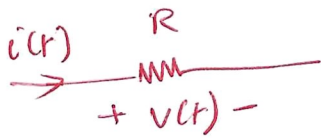
$$\frac{30s + 10}{s^2 + 9s + 20} = \frac{30s + 10}{(s+4)(s+5)} = \frac{-20}{s+4} + \frac{50}{s+5}$$

$$F(s) = s^2 + 4s + 10 + \frac{20}{s+4} + \frac{50}{s+5}$$

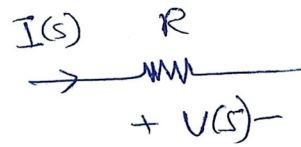
$$f(t) = \frac{d^2}{dt^2} \delta(t) + 4 \frac{d}{dt} \delta(t) + 10 \delta(t) + (20 e^{-4t} - 50 e^{-5t}) u(t)$$

Circuit element models Chapter 13g-

* Resistor



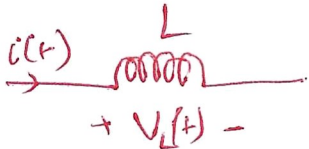
$$v(t) = R i(t)$$



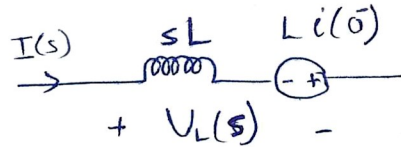
$$V(s) = R I(s)$$

$$Z_R = R \Omega$$

* Inductor



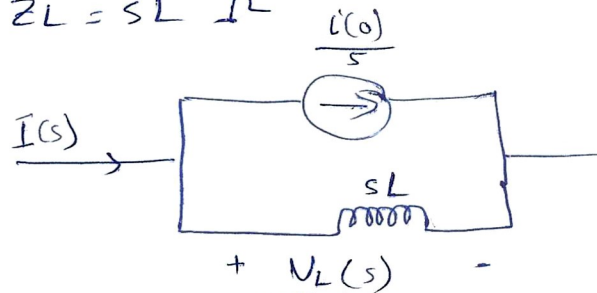
$$v_L(t) = L \frac{di}{dt}$$



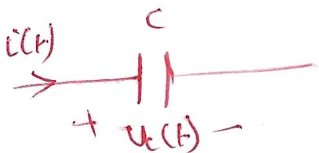
$$V_L(s) = L [sI(s) - i(0)]$$

$$= sL I(s) - L i(0)$$

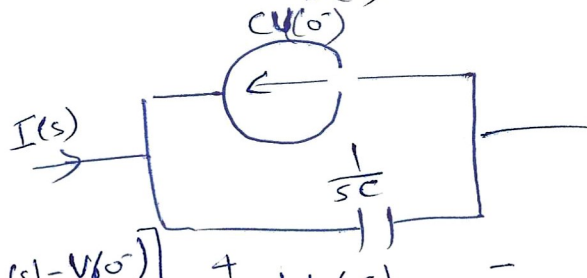
$$Z_L = sL \Omega$$



* Capacitor



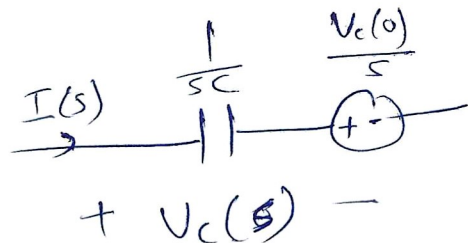
$$i(t) = C \frac{dv_C(t)}{dt}$$



$$I(s) = C [s v_C(s) - v_C(0)] + v_C(s)$$

$$= sC v_C(s) - C v_C(0)$$

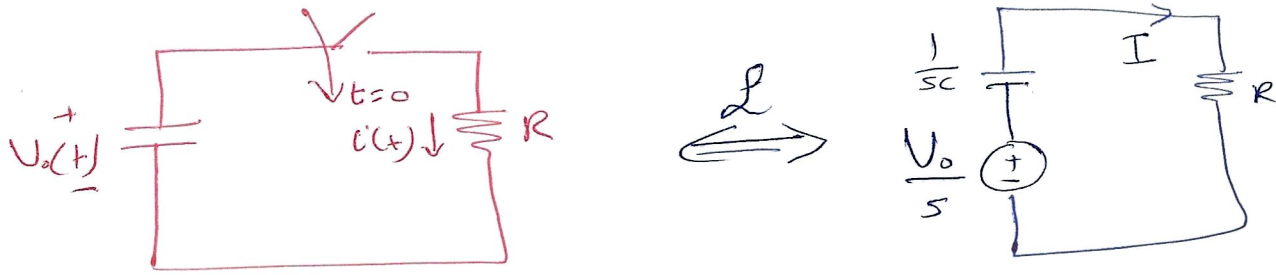
$$Z_C = \frac{1}{sC} \Omega$$



Natural response of an RC circuit

Natural response - No active source in the circuit

find $i(t)$ and $v(t)$



KVL

$$-\frac{V_0}{s} + I\left(\frac{1}{sC}\right) + IR = 0$$

$$I\left(\frac{1}{sC} + R\right) = \frac{V_0}{s} \Rightarrow I = \frac{\frac{V_0}{s}}{\frac{1}{sC} + R} = \frac{V_0}{\frac{1}{C} + Rs}$$

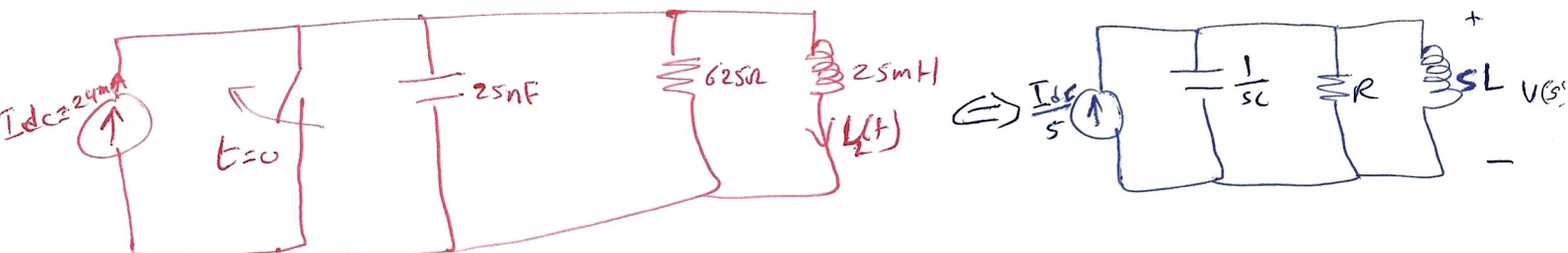
$$I(s) = \frac{V_0/R}{s + \frac{1}{RC}}$$

$$\therefore i(t) = \frac{V_0}{R} e^{-\frac{1}{RC}t} u(t)$$

$$v(t) = V_0 e^{-\frac{1}{RC}t} u(t)$$

step response of a parallel RLC circuit

step response - active source is applied to the circuit



$$KCL \quad \frac{I_{dc}}{s} = \frac{V(s)}{\frac{1}{sC}} + \frac{V(s)}{R} + \frac{V(s)}{sL} = V\left(sC + \frac{1}{R} + \frac{1}{sL}\right)$$

$$V = \frac{\frac{I_{dc}}{s}}{sC + \frac{1}{R} + \frac{1}{sL}} = \frac{I_{dc}}{s^2C + \frac{s}{R} + \frac{1}{L}} = \frac{I_{dc}/C}{s^2 + \frac{s}{RC} + \frac{1}{LC}}$$

$$I_L = \frac{V}{sL} = \frac{I_{dc}/C}{s^2 + \frac{s}{RC} + \frac{1}{LC}} \cdot \frac{1}{sL}$$

$$= \frac{I_{dc}/CL}{s(s^2 + \frac{s}{RC} + \frac{1}{LC})}$$

$$= \frac{3.84 \times 10^7}{s(s^2 + (6.4 \times 10^4)s + (1.6 \times 10^9))}$$

$$= \frac{24}{s} + \frac{20 \sqrt{127}}{s - (-32k + j24k)} + \frac{20 \sqrt{-127}}{s - (-32k - j24k)} \text{ mA.s}$$

$$i_L(t) = 24 u(t) + 20 e^{j127t} e^{-(32k)t} e^{j(24k)t} u(t) + 20 e^{-j127t} e^{-(32k)t} e^{-j(24k)t} u(t)$$

$$= [24 + 40 e^{-32kt} \cos(24kt + 127t)] u(t) \text{ mA}$$

In the previous example if $i(t) = I_m \cos(\omega t) u(t)$

$$\therefore I(s) = I_m \frac{s}{s^2 + \omega^2}$$

$$\Rightarrow V(s) = \frac{(I_m/C) s^2}{(s^2 + \omega^2)(s^2 + \frac{s}{RC} + \frac{1}{LC})}$$

$$I(s) = \frac{V(s)}{sL} = \frac{(I_m/CL) s}{(s^2 + \omega^2)(s^2 + \frac{s}{RC} + \frac{1}{LC})}$$

$$= \frac{K_1}{s - j\omega} + \frac{K_1^*}{s + j\omega} + \frac{K_2}{s - (\alpha + j\beta)} + \frac{K_2^*}{s - (-\alpha - j\beta)}$$

$$i(t) = \underbrace{2|K_1| \cos(\omega t + \angle K_1)}_{\text{steady}} + \underbrace{2|K_2| e^{-\alpha t} \cos(\beta t + \angle K_2)}_{\text{transient}}$$