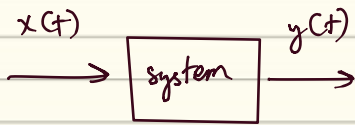


**System Modeling in the time domain**  
**By: Rawan Alfares**



# system modeling in the time domain



## Properties :-

1] Continuous time and Discrete time.

2] Time-invariant and Time variant Systems.

ex: if  $y(t) = x(t)$ , check if  $y(t)$  is fixed or not fixed.

1. delay for time input, "function of time".

$$y_1(t-t_0) = x_1(t-t_0) \rightarrow (1)$$

2. Time delay

$$y_2(t-t_0) = x_2(t-t_0) \rightarrow (2)$$

3. Compare eq<sub>1</sub> and eq<sub>2</sub>

$y_1 \neq y_2$ , so Time variant system. ["يغير مع الزمن"]

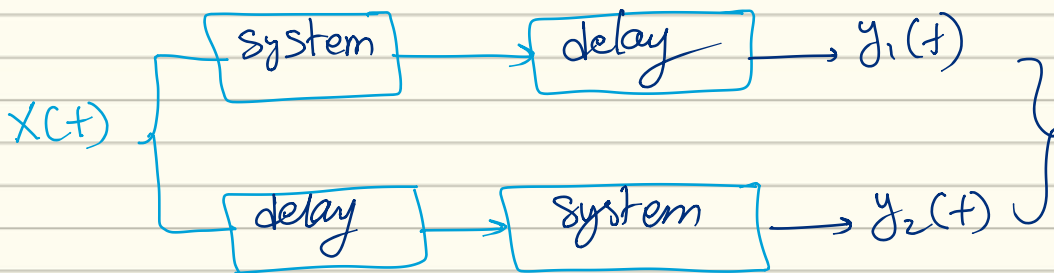
ex2:  $y(t) = x(t)$

$$y_1(t-t_0) = x_1(t-t_0) \rightarrow (1)$$

$$y_2(t-t_0) = x_2(t-t_0) \rightarrow (2)$$

$y_1 = y_2$ , so Time invariant,

"no delay" "output"



if  $y_1(t) = y_2(t)$   
time invariant

if  $y_1(t) \neq y_2(t)$   
time variant.

ex3:  $y(t) = x(t-5) + x(3-t)$

$$y_1(t) = x(t-5-t_0) + x(3-t-t_0)$$

$$y_2(t) = x(t-t_0-5) + x(3-t+t_0)$$

$y_1(t) \neq y_2(t)$ , Time variant.



ex.  $y(t) = \sin(x(t)) \rightarrow$  time invariant

$y(t) = x(t^2) \rightarrow$  time variant.

### 3] Causal and non-causal

past or current values "قيم الحاضر أو الوقت الحاضر"  
 future values "قيم المستقبل"

output values > input values

ex. if the system  $3y(t) + \int_{-\infty}^t y(\tau) d\tau = x(t)$

تتعلق القيمة الحالية "output" بالقيمة الموجودة في الماضي  
 تعتمد على قيمة الماضي

Causal.

ex. The system  $y(t) = x(t^2)$    
 $\left. \begin{array}{l} 0 \leq t \leq 1 \\ t > 1 \end{array} \right\} \begin{array}{l} \text{Causal system} \\ \text{non-causal system} \end{array}$

\* In general, this system is non-causal.

ex.  $y(t) = 10x(t+2) + 5$

future value  $\rightarrow$  so, non-causal system.

since values depends on future.

### 4] memory & memory less "Dynamic and instantaneous"

use buffer

\* values between brackets in input should equal values between brackets in output.

memory less

ex.  $y(t) = x(t)$ , memory less, "Instantaneous".

ex.  $2 \frac{dy(t)}{dt} + 3y(t) = \frac{d^2x(t)}{dt^2} + x(t)$ , memory/dynamic

differentiable equation  $\Rightarrow$  "لا تأملها، تصبح على قرة زمنية وليست على زمن محدد"

ex.  $y(t+2) = y(t+2) \rightarrow$  memory less.

•  $y(t) = x(t^2)$

•  $y(t) = x(t+1)$

} memory



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\* لو كانت مشتقة أو تأمل دائماً "memory dynamic".

# 5] linear & non linear Systems

linear & Time invariant.

linear  $\left\{ \begin{array}{l} \text{Scaling} \\ \text{Adding} \end{array} \right.$

ex.  $y(t) = 3x(t)$

linear  $\left\{ \begin{array}{l} \text{Constant} \\ \text{لا} \end{array} \right.$ , Constant

Constant  $\left\{ \begin{array}{l} \text{لا} \\ \text{لا} \end{array} \right.$   $\neq$

ex.  $y(t) = 3x(t) + 2$ . non linear

$y_1(t) = \alpha 3x_1(t)$

$y_2(t) = \alpha 3x_2(t)$

$y_3 = y_1 + y_2$

$y_3 = 3\alpha x_1(t) + 3\alpha x_2(t)$

$y_4(t) = \alpha [3x_1(t) + 3x_2(t)]$

$y_4(t) = 3\alpha x_1(t) + 3\alpha x_2(t)$

$\alpha \cdot y_1$  1  
 $\alpha \cdot y_2$  2  
 $\alpha x_1 + \alpha x_2$  3  
 $y_3 = \alpha y_1 + \alpha y_2$  3  
 $x_1 + x_2 = y_4$  4

$y_3(t) = y_4(t)$   
 so, linear

$\alpha y_1(t) = 3\alpha x_1(t) + 2\alpha$

$\alpha y_2(t) = 3\alpha x_2(t) + 2\alpha$

$y_3(t) = 3\alpha x_1(t) + 3\alpha x_2(t) + 4\alpha$

$y_4(t) = 3 [\alpha x_1(t) + \alpha x_2(t)] + 2$

$y_4(t) = 3\alpha x_1(t) + 3\alpha x_2(t) + 2$

$y_3 \neq y_4$

ex.  $\frac{dy}{dt} + 3y(t) = 2x(t)$

$\alpha y_1' + 3\alpha y_1(t) = \alpha 2x_1(t)$

$\alpha y_2' + 3\alpha y_2(t) = \alpha 2x_2(t)$

$y_3(t) = \alpha [y_1' + y_2'] + 3\alpha [y_1 + y_2] = 2\alpha [x_1(t) + x_2(t)]$

$y_4(t) = \alpha [y_1' + y_2'] + 3\alpha [y_1 + y_2] = 2\alpha [x_1(t) + x_2(t)]$

$y_3 = y_4$   
 linear system

ex.  $\frac{dy}{dt} + 3y(t) + 5 = 2x(t)$

$\alpha y_1' + \alpha 3y_1(t) + \alpha 5 = \alpha 2x_1(t)$

$\alpha y_2' + \alpha 3y_2(t) + \alpha 5 = \alpha 2x_2(t)$

$y_3(t) = \alpha [y_1' + y_2'] + 3\alpha [y_1 + y_2] + 10\alpha = 2\alpha [x_1 + x_2]$

$y_4(t) = \alpha [y_1' + y_2'] + 3\alpha [y_1 + y_2] + 5 = 2\alpha [x_1(t) + x_2(t)]$

$y_3 \neq y_4$ ,  
 so, non linear system.

ex.  $y(t) = \int_{-\infty}^t x(\tau) d\tau$

$y_3 = \alpha \int_{-\infty}^t x_1(\tau) d\tau + \alpha \int_{-\infty}^t x_2(\tau) d\tau$

$y_4(t) = \alpha \int_{-\infty}^t (x_1(\tau) + x_2(\tau)) d\tau$

$y_3 = y_4$

so, linear.



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ex.  $y[n] = x^2[n]$

$y_3[n] = \alpha x_1^2[n] + \alpha x_2^2[n]$   
 $y_4[n] = \alpha (x_1[n] + x_2[n])^2$

$y_3 \neq y_4$   
 non linear system.

ex. which of the following signals is linear, causal, time invariant and memory, justify your answer.

1)  $y(t) = x(t-2) + x(2-t)$

A) linearity :-  
 $\alpha y_1(t) = \alpha x_1(t-2) + \alpha x_1(2-t)$   
 $\alpha y_2(t) = \alpha x_2(t-2) + \alpha x_2(2-t)$

$y_3(t) = \alpha [x_1(t-2) + x_2(t-2)] + \alpha [x_1(2-t) + x_2(2-t)]$   
 $y_4(t) = \alpha [x_1(t-2) + x_2(t-2)] + \alpha [x_1(2-t) + x_2(2-t)]$

$y_3 = y_4$   
 linear system

B) Causality :- when  $t=0$   
 $y(0) = x(-2) + x(2)$   
 present past future  $\rightarrow$  so, non causal.

C) Time invariance :-  
 $y_1(t) = x(t-2-t_0) + x(2-t-t_0)$   
 $y_2(t-t_0) = x(t-t_0-2) + x(2-(t-t_0))$   
 $y_2(t-t_0) = x(t-t_0-2) + x(2-t+t_0)$

time variant  
 $y_1 \neq y_2$


D) memory :- memory,  $x(0) = x(-2) + x(2)$

2)  $y(t) = \cos(3t)x(t)$

1) linearity :-  
 $y_3 = \alpha \cos(3t)x_1(t) + \alpha \cos(3t)x_2(t)$   
 $y_4 = \alpha \cos 3t [x_1(t) + x_2(t)]$   
 $y_4 = \alpha \cos 3t x_1(t) + \alpha \cos 3t x_2(t)$

$y_3 = y_4$   
 linear.

2) Causality :-  
 $y(t) = \cos(3t)x(t)$   
 present  
 so causal.

  
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 constant  
 y, x

3) Time invariant :  $y_1(t) = \cos(3t - t_0) x(t - t_0)$

$$y_2(t - t_0) = \cos(3t - 3t_0) x(t - t_0)$$

time variant

4) memory :- memory less ,  $y(t) = \cos(3t) x(t)$

same input  
between parentheses.

$$3) y(t) = \int_{-\infty}^{2t} x(\tau) d\tau.$$

1) linear :  $y_1(t) = \alpha \int_{-\infty}^{2t} x_1(\tau) d\tau.$

$$y_2(t) = \alpha \int_{-\infty}^{2t} x_2(\tau) d\tau$$

$$y_3(t) = \alpha \int_{-\infty}^{2t} x_1(\tau) d\tau + \alpha \int_{-\infty}^{2t} x_2(\tau) d\tau.$$

$$y_4(t) = \alpha \int_{-\infty}^{2t} (x_1(\tau) + x_2(\tau)) d\tau.$$

$y_3 = y_4$   
So, linear  
system.

2) Causality :- non Causal. ,  $(t)$  depends on future values ( $2t$ ) value.

3) Time - variant :-  $y_1(t) = \int_{-\infty}^{2t - t_0} x(\tau) d\tau.$

$$y_2(t) = \int_{-\infty}^{2t - 2t_0} x(\tau) d\tau.$$

Time variant  
system

4) memory :- memory system.



4)  $y(t) = x(t-5) + x(3-t)$

\* Linearity :-  $y_3(t) = \alpha x_1(t-5) + \alpha x_1(t-5) + \alpha x_2(t-5) + \alpha x_2(3-t)$

$y_4(t) = \alpha (x_1(t-5) + x_1(3-t)) + \alpha (x_2(t-5) + x_2(3-t))$

So, linear system.

\* Causality :-  $y(0) = y(-5) + x(3)$  , non Causal system.   
 *future*

\* memory :- , memory system.

\* Time variant :-  $y_1(t) = x(t-5-t_0) + x(3-t-t_0)$

$y_2(t) = x(t-t_0-5) + x(3-(t-t_0))$

$y_3(t) = x(t-t_0-5) + x(3-t+t_0)$

$y_1 \neq y_2$   
time variant..

6) Stable and unstable

bounded input  $\rightarrow$  [ ]  $\rightarrow$  bounded  $\Rightarrow$  stable  $\Rightarrow$  (finite).

unbounded  $\Rightarrow$  unstable  $\Rightarrow$  goes to infinity.

$|x(t)| \leq M, \forall t$

ex.  $y(t) = e^{x(t)}$   
 $= e^M \rightarrow$  constant, so stable.

ex.  $x(t) e^{2t}$  unstable.

ex.  $x(t) e^2$  stable

ex.  $x(t) e^{-2t}$  stable

ex.  $y(t) = \int_{-\infty}^t x(z) dz$  unstable

\* bounded means Can reach max and min value.   
 \* it doesn't go to  $\infty$ .

-: function  $x(t) = u(t)$



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$\rightarrow \int u(t) = r(t)$ , as  $t \rightarrow \infty, r(t) \rightarrow \infty$   
So, it's unbounded  
then the system is unstable

ex.  $y(t) = 2x^2(t)$

$x(t) = u(t) \rightarrow y(t) = 2u^2(t)$

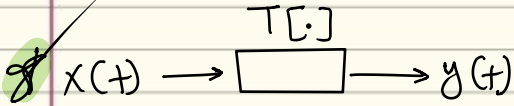
$t=1, \rightarrow y(1) = 2$ , fixed, bounded.

so the system is stable.

6) Invertible and non Invertible

هل يقدم أكتب input بخواص output

$(1-1) \rightarrow$  invertible  
 if output of input is  
 $\neq$  not  $(1-1) \rightarrow$  non invertible  
 more than one input  
 give the same output



$T[x(t)] = y(t)$

$x(t) = T^{-1}[y(t)]$

ex.  $y(t) = 10x(t) + 3$

$T^{-1} = x(t) = \frac{y(t) - 3}{10} \rightarrow$  so the system is invertible

ex.  $y[n] = x^2[n]$  non invertible

ex.  $y(t) = \int_{-\infty}^t x(\tau) d\tau$  invertible

$T^{-1} = x(t) = y'(t)$

Invertible  $\rightarrow$  هل يقدم أكتب input بخواص output

function  $T^{-1}$  أووجد

output of input  $\neq$  unique input, unique output





**Examples** Which of the following signals is Linear, Causal memory?

1.  $y(t) = x(t-2) + x(2-t)$

2.  $y(t) = x\left(\frac{t}{2}\right)$

3.  $y(t) = \begin{cases} 0, & t < 0 \\ x(t) + x(t-2), & t \geq 0 \end{cases}$

4.  $y(t) = \begin{cases} 0, & x(t) < 0 \\ x(t) + x(t-2), & x(t) \geq 0 \end{cases}$

	linearity	Causality	Time invariant	memory
1.	linear	non Causal	Time variant	memory
2.	linear	non-Causal	Time variant	memory
3.	linear	Causal	Time variant	memory
4.	non Linear	Causal	Time invariant	memory

$$y(t) = \ln |x(2^t)|$$

↳ causal, non causal.



# LTI System Linear Time Invariant

$$x(t) \longrightarrow \boxed{h(t)} \longrightarrow y(t)$$

When it is LTI system we always do Convolution Integral.

$$y(t) = x(t) * h(t)$$

$$* y(t) = \int_{-\infty}^{\infty} x(t-\tau) h(\tau) d\tau$$

$$* \text{or } y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

$$* \int_{-\infty}^{\infty} x(t) h(k-t) dt$$

$$* \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

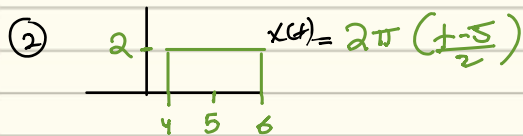
$$* \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

$$y[n] = \sum_{n=-\infty}^{\infty} x[n] h[n-n]$$

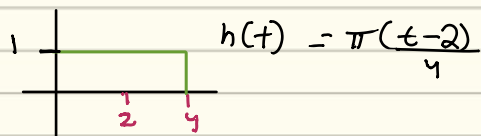
ex. for LTI system if

$$x(t) = 2\pi \frac{(t-5)}{2} \text{ and } h(t) = \pi \frac{(t-2)}{4}$$

Ans: ① LTI system, evaluate  $y(t)$ ,  $y(t) = h(t) * x(t)$



$[4, 6] \leftarrow$  Range



$[0, 4] \leftarrow$  Range

③  $[4, 6]$  جمع كل رقم في الفترة  
 $[0, 4]$  الأرقام مع الأرقام في الفترة

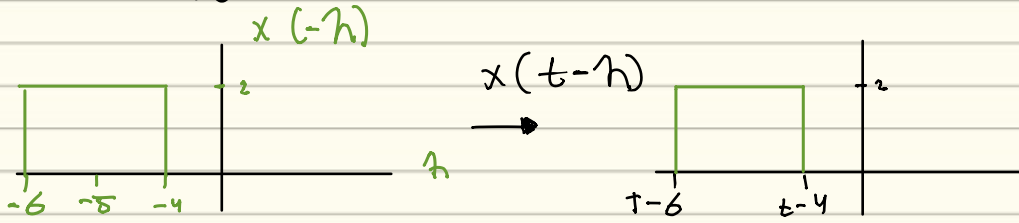
$[4, 8, 6, 10]$  الثانية  
 ته يرتفع  
 تصاعدياً

$[4, 6, 8, 10]$  ← الفترة التي  
 راع مشتق عليها

lecture 11



$$y(t) = \int_{-\infty}^{\infty} x(t-\tau) h(\tau) d\tau$$



for  $-\infty < t < 4$

$-\infty - 6 < t$  |  $4 - 4 < 0$

لا يوجد تقاطع

$y(t) = 0$

$4 < t < 6$

$y(t) = \int_0^{t-4} (1)(2) d\tau$

$= 2(t-4)$

$6 < t < 8$

$y(t) = \int_{t-6}^{t-4} (1)(2) d\tau$

$= 2(t-4 - t+6)$

$= 4$

$8 < t < 10$

$y(t) = \int_{t-6}^4 (1)(2) d\tau$

$= 2(-2+10)$

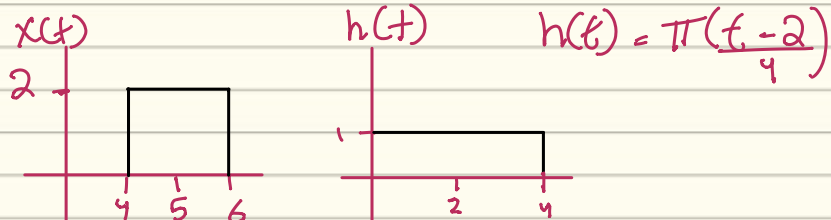
$t > 10$

لا يوجد تقاطع

$y(t) = 0$



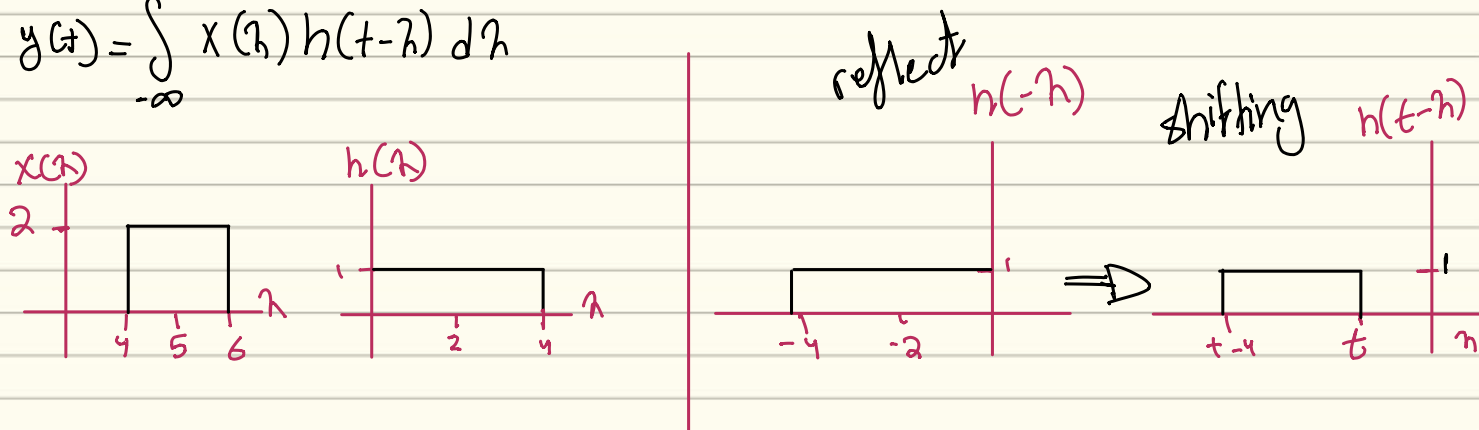
another way to solve ex.  $X(t) = 2\pi \left( \frac{t-5}{2} \right)$



$[4, 6]$   $[2, 4]$

Rang:  $[4, 6, 8, 10] \rightarrow (-\infty, 4, 6, 8, 10, \infty)$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

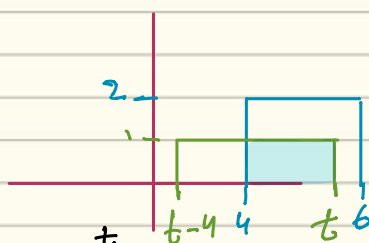


1)  $-\infty < t < 4$



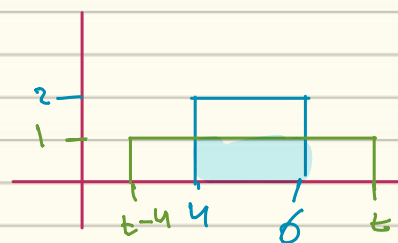
$$y(t) = \int_{-\infty}^4 0 d\tau = 0$$

2)  $4 < t < 6$



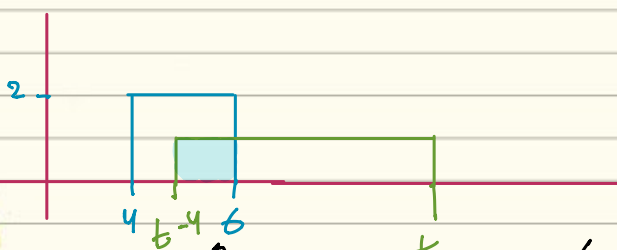
$$y(t) = \int_4^t (1)(2) d\tau = 2\tau \Big|_4^t = 2t - 8$$

3)  $6 < t < 8$



$$y(t) = \int_4^6 (1)(2) d\tau = 2\tau \Big|_4^6 = 4$$

4)  $8 < t < 10$



$$y(t) = \int_{t-6}^6 (1)(2) d\tau = 2\tau \Big|_{t-6}^6 = 2(6-t+4)$$



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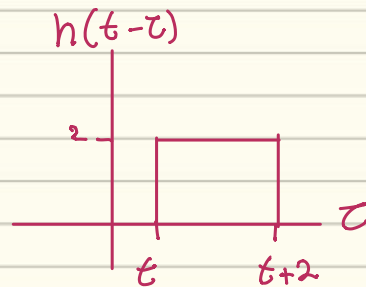
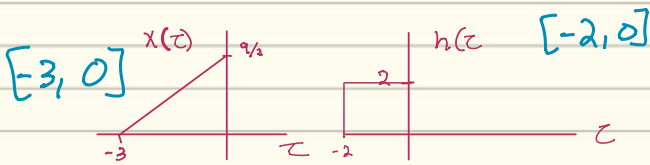
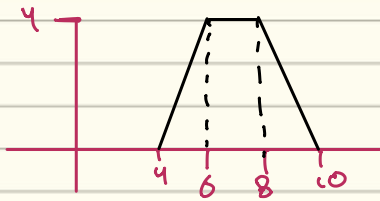
5)  $10 < t < \infty$



$$y(t) = \int_0^{\infty} 0 \, dh = 0$$

$$y(t) = 0, \quad -\infty < t < 4$$

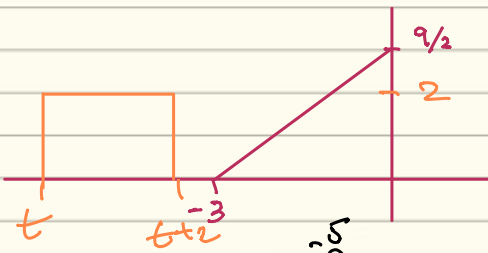
$$\left. \begin{aligned} &2t-8, \quad 4 < t < 6 \\ &1, \quad 6 < t < 8 \\ &20-2t, \quad 8 < t < 10 \\ &0, \quad t > 10 \end{aligned} \right\}$$



$$y(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau) \, d\tau$$

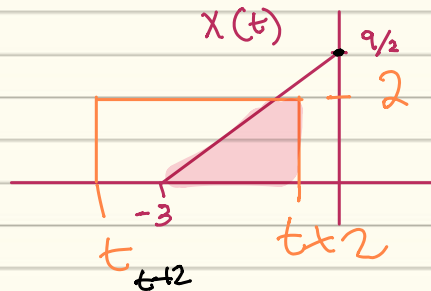
$[-5, -3, -2, 0]$

6)  $-\infty < t < -5$



$$y(t) = \int_{-\infty}^{\infty} 0 \, d\tau = 0$$

7)  $-5 < t < -3$



$$m = \frac{9/2 - 0}{0 - (-3)} = \frac{9/2}{3} = \frac{3}{2}$$

$$y - 0 = \frac{3}{2}(x + 3)$$

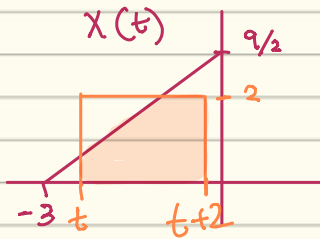
$$x(t) = \frac{3}{2}(t + 3)$$

$$\begin{aligned} y(t) &= \int_{-3}^{t+2} (2) \left( \frac{3}{2}(\tau + 3) \right) \, d\tau \\ &= \left( \frac{3\tau^2}{2} + 9\tau \right) \Big|_{-3}^{t+2} \end{aligned}$$



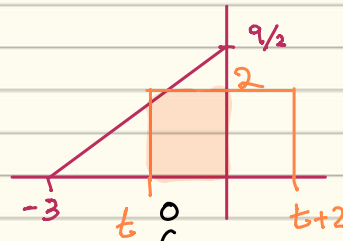
$$= \frac{3(t+2)^2}{2} + 9(t+2) - \left( \frac{3(9)}{2} + 9(-3) \right)$$

$$\text{[3]} -3 < t < -2$$



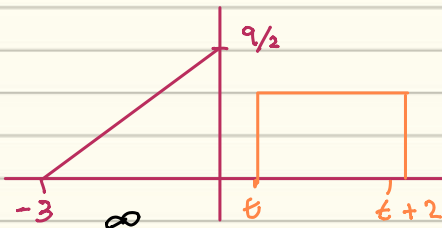
$$y(t) = \int_t^{t+2} (2) \left( \frac{3}{2}(\tau+3) \right) d\tau$$

$$\text{[4]} -2 < t < 0$$



$$y(t) = \int_t^{t+2} (2) \left( \frac{3}{2}(\tau+3) \right) d\tau$$

$$\text{[5]} 0 < t$$



$$y(t) = \int_0^{\infty} 0 d\tau = 0.$$

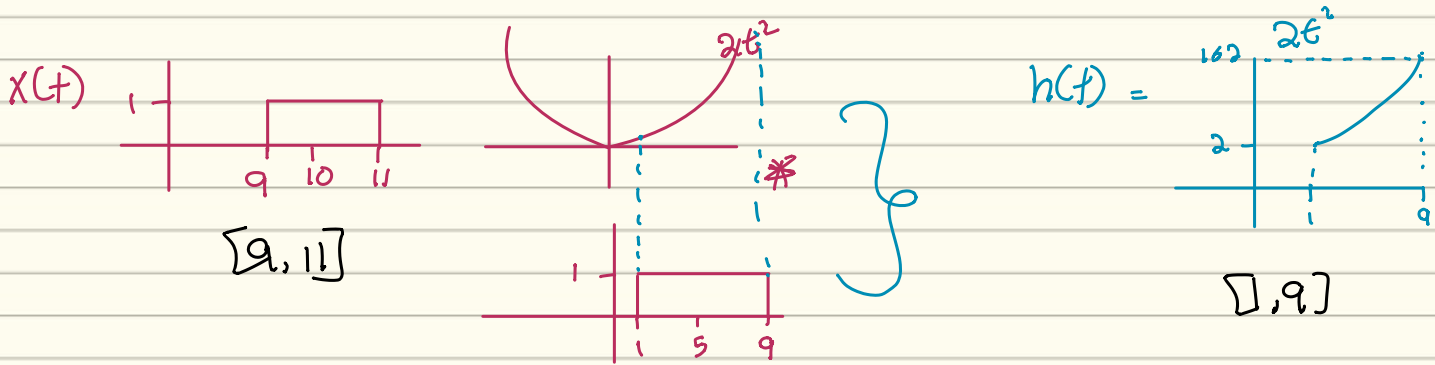


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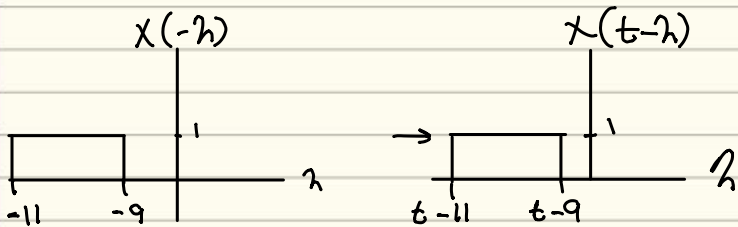
Example 2.7: Compute, Using the convolution integral, the response of the LTI with impulse response

$$h(t) = 2t^2 \Pi\left(\frac{t-5}{8}\right) \text{ to the input } x(t) = \Pi\left(\frac{t-10}{2}\right)$$

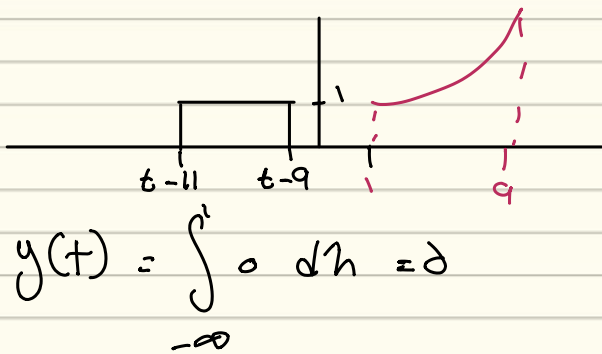


$$y(t) = x(t) * h(t)$$

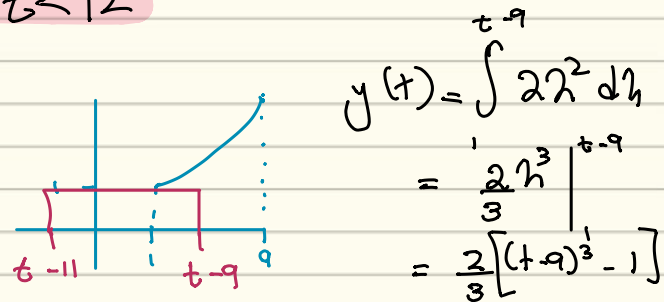
$$= \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau, \text{ Range } [10, 12, 18, 20]$$



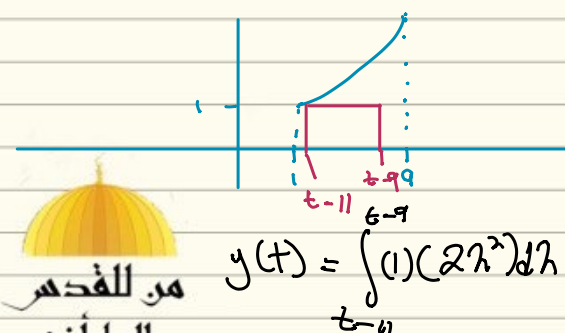
①  $10 < t$



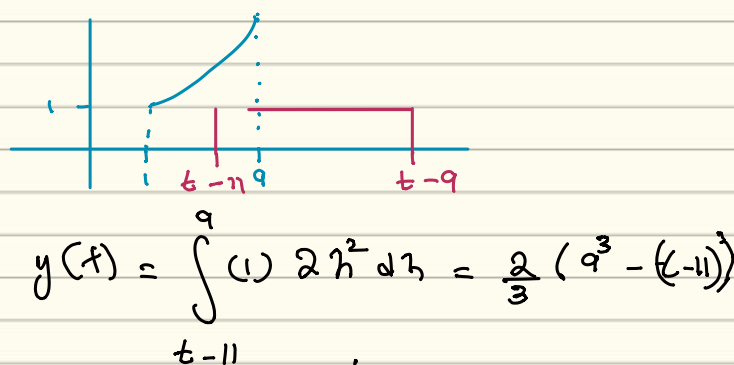
②  $10 < t < 12$



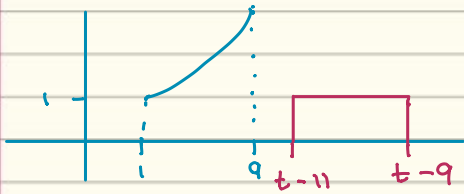
③  $12 < t < 18$



④  $18 < t < 20$



⑤  $20 < t$

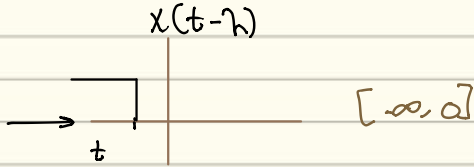
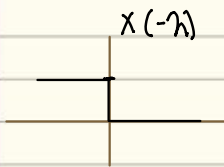
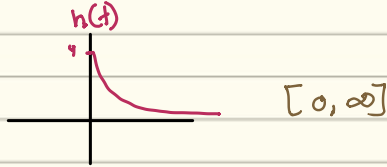
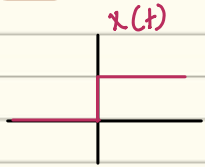


$$y(t) = \begin{cases} 0, & t < 10 \\ \frac{2}{3}[(t-9)^3 - 1], & 10 \leq t < 12 \\ \frac{2}{3}[(t-9)^3 - (t-10)^3], & 12 \leq t < 18 \\ \frac{2}{3}[9^3 - (t-11)^3], & 18 \leq t < 20 \\ 0, & t \geq 20 \end{cases}$$

$$y(t) = \int_{20}^{\infty} 0 \, d\tau = 0$$

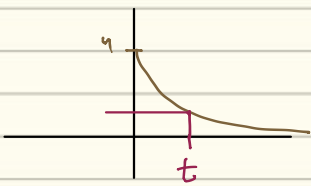
الزمن نعوطن [10, 12, 18, 20]  
في  
نقطتهم.

ex.  $x(t) = 2u(t)$ ,  $h(t) = 4e^{-3t}u(t)$ .



الفترة الكلية  $[0, \infty]$

$0 < t < \infty$



$$y(t) = \int_0^t (2)(4)(e^{-3\tau}) \, d\tau$$

$$= \frac{8 \cdot e^{-3\tau}}{-3} \Big|_0^t = -\frac{8}{3}e^{-3t} + \frac{8}{3}$$



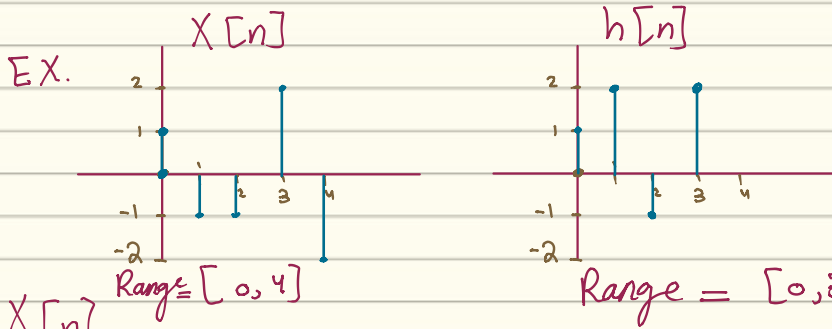
من للفهد

إلى أفند



# Discrete

$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$



$X[n]$ $h[n]$	1	-1	-1	2	-2
1	1	-1	-1	2	-2
2	2	-2	-2	4	-4
-1	-1	1	1	-2	2
2	2	-2	-2	4	-4

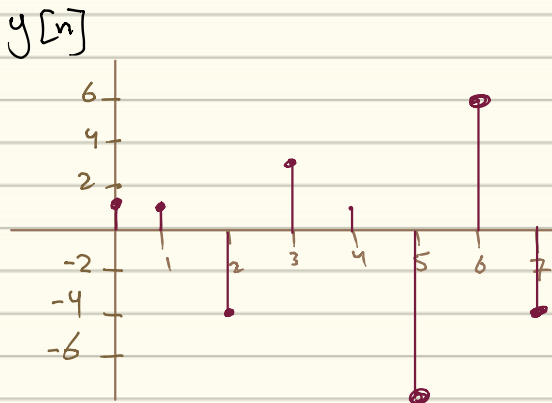
The value of each time in  $x[n]$

≠ each value in  $x[n]$ , multiply it by each value in  $h[n]$

values of time in  $h[n]$

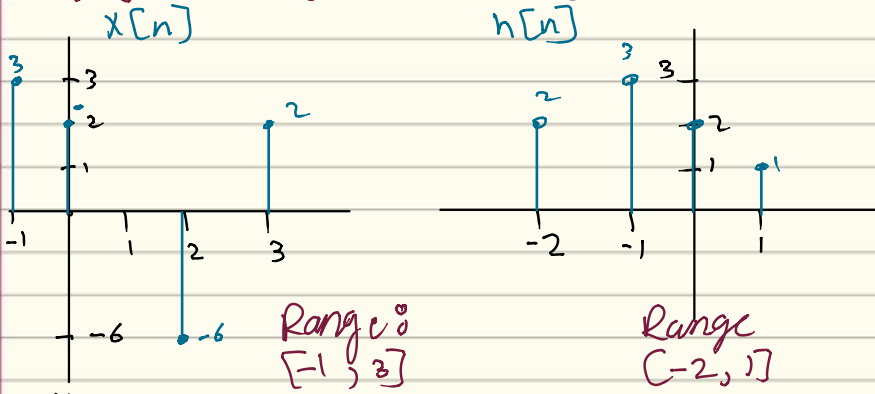
$$y[n] = \{1, 1, -4, 3, 1, -8, 6, -4\}$$

$n = \{0, 1, 2, 3, 4, 5, 6, 7\}$



Ex.  $x[n] = 3\delta[n+1] + 2\delta[n] - 6\delta[n-2] + 2\delta[n-3]$

$h[n] = \delta[n-1] + 2\delta[n] + 3\delta[n+1] + 2\delta[n+2]$



total Range  
 $[-1, 3] \cup [-2, 1]$

$[-3, 0, 1, 4]$   
 $\Downarrow$   
 $[-3, 4]$

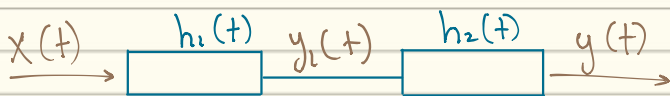
$x[n]$	3	2	0	-6	2
$h[n]$					
2	6	4	0	-12	4
3	9	6	0	-18	6
2	6	4	0	-12	4
1	3	2	0	-6	2

$y[n] = \{ 6, 13, 12, -5, -12, -6, -2, 2 \}$   
 $n = -3 \quad -2 \quad -1 \quad 0 \quad 1 \quad 2 \quad 3 \quad 4$

**Properties of convolution**

1) Commutative (تبادلية)  $\xrightarrow{x(t)} \boxed{\text{LTI}} \xrightarrow{h(t)} y(t) = x(t) * h(t) = h(t) * x(t) = \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$

2) Associative (تجميعية)

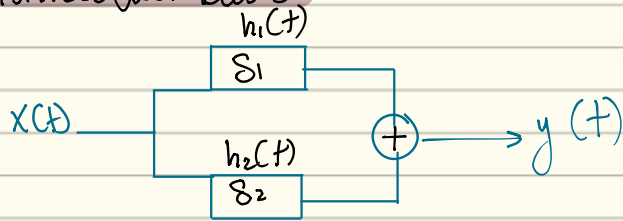


$[x(t) * h_1(t)] * h_2(t) = y(t)$

$x(t) * [h_1(t) * h_2(t)] = y(t)$



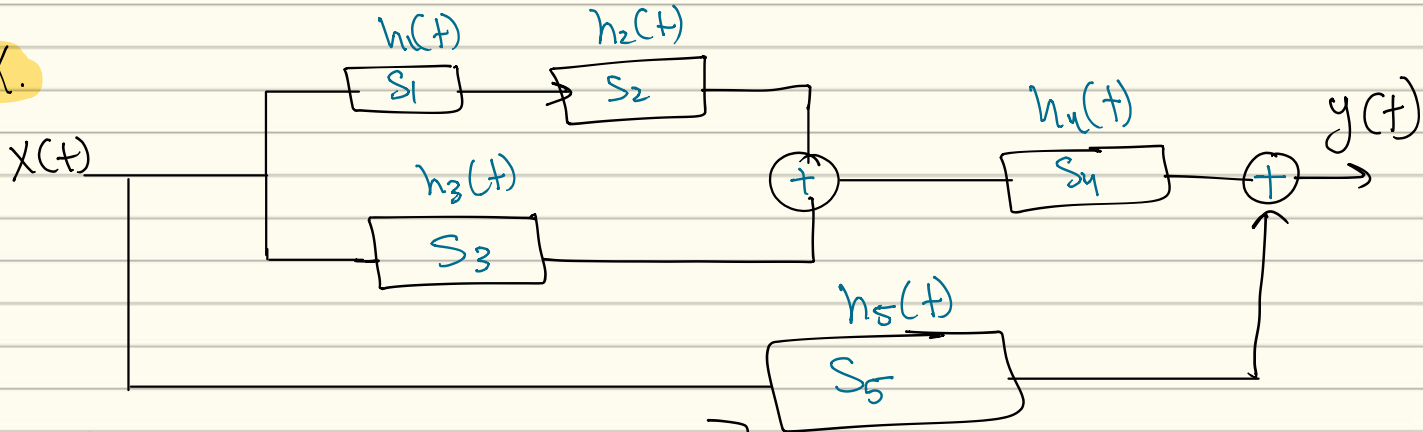
### 3) Distributive law :-



$$x(t) * [h_1(t) + h_2(t)] = y(t)$$

$$= x(t) * h_1(t) + x(t) * h_2(t) = y(t)$$

ex.



$$\Rightarrow \left[ \left[ (h_1(t) * h_2(t)) + h_3(t) \right] * h_4(t) \right] + h_5(t) = y(t)$$

### Summary of properties :-

① Commutative :-  $x_1 * x_2 = x_2 * x_1$

② Distributive :-  $x_1 * (x_2 + x_3) = x_1 * x_2 + x_1 * x_3$

③ Associative :-  $(x_1 * x_2) * x_3 = x_1 * (x_2 * x_3)$

④ Shifting property :- if  $x_1 * x_2 = z(t)$

then  $x_1(t) * x_2(t-t_0) = z(t-t_0)$

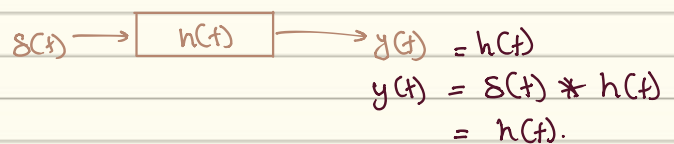
$x_1(t-t_0) * x_2(t) = z(t-t_0)$

$x_1(t-t_1) * x_2(t-t_2) = z(t-t_1-t_2)$

$= z(t - [t_1 + t_2])$

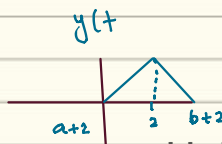
### ⑤ Convolution with impulse function :-

①  $x(t) * \delta(t) = x(t)$



②  $x(t) * \delta(t-t_0) = x(t-t_0)$

ex.  $x(t) * \delta(t-2) = x(t-2)$



\* أي شيء نضربها (convolution) مع دلتا (delta function) يعطينا الفنتكشن نفسه



EX.  $x(t) \longrightarrow \boxed{h(t)} \longrightarrow y(t)$

Where  $h(t) = \delta(3t-4)$ . Find  $y(t)$  in terms of  $x(t)$ .

Ans 8-  $y(t) = x(t) * h(t)$   
 $= x(t) * [\delta(3t-4)]$   
 $= x(t) * \frac{1}{3} \delta(t - \frac{4}{3})$   
 $= \frac{1}{3} x(t - \frac{4}{3})$

$\delta(3t-4) = \delta(3(t - \frac{4}{3}))$   
 $= \frac{1}{3} \delta(t - \frac{4}{3})$

Ex. let the impulse response of the continuous time system be.

$h(t) = \delta(t-1) + \delta(t-3)$

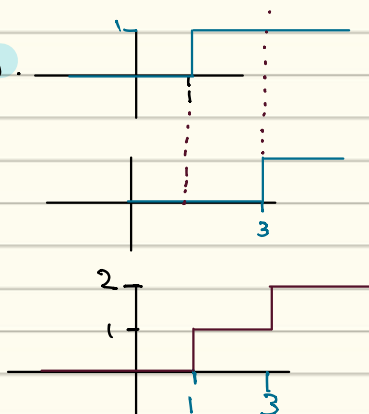
- ① Find the step response at  $t=5$  sec.
- ② Find the step response at  $t=2$  sec.

□  $x(t) = u(t) \longrightarrow \boxed{h(t)} \longrightarrow y(t)$

Ans 8-  $y(t) = x(t) * h(t)$   
 $= u(t) * (\delta(t-1) + \delta(t-3))$  *by distributive law*  
 $= u(t) * \delta(t-1) + u(t) * \delta(t-3)$   
 $= u(t-1) + u(t-3)$  , at  $t=5$ ,  $y(5) = u(4) + u(2)$   
 $= 1 + 1 = 2$

at  $t=2$ .  $y(2) = u(1) + u(-1)$   
 $= 1 + 0$   
 $= 1$

Plot  $y(t)$ .



**NOTE 8-**

\* impulse response

input  $\delta(t)$  find output.

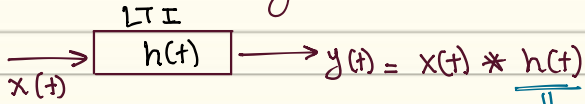
\* Step response

input  $u(t)$  find output



# properties of continuous LTI system

We said that any input  $x(t)$ , we can find the output by convolution  $x(t)$  with  $y(t)$ .



منه قاعة وفيه نلاحظ بوجود  $y(t)$  وبالتالي، من خلاله نجد خصائصه (system) اذ  
Causal memory stable

## 1] System with memory .

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

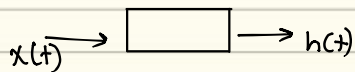
$y(t)$  depends on  $x(\tau)$  for  $\tau \in (-\infty, \infty)$ , In general, the system have memory.

\* only one special case to memory less system.

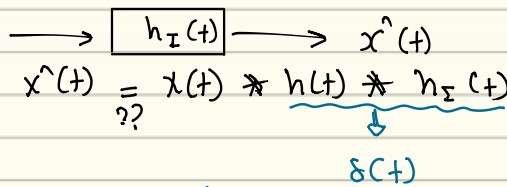
$h(t) = k \delta(t)$  }  $y(t) = \int_{-\infty}^{\infty} x(\tau) k \delta(t-\tau) d\tau = k x(t)$   
Constant

## 2] Invertability :-

$$y(t) = x(t) * h(t)$$



$$\hat{x}(t) = y(t) * h_I(t)$$



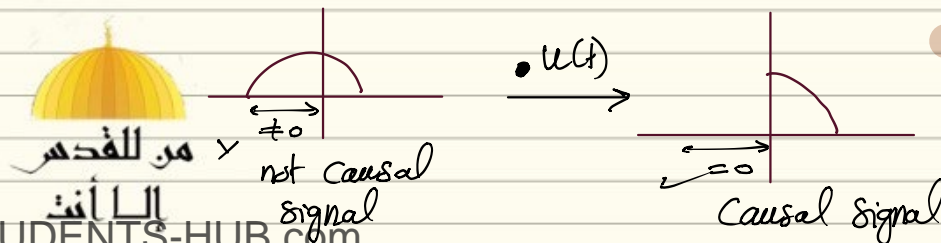
LT I is invertable iff  $\exists h_I(t)$  such that  $h_I(t) * h(t) = \delta(t)$  → We will see that on ch.4 Fourier transform

## 3] Causality :-

claim :- let  $h(t)$  be causal signal.

\* Causal signal  $\Rightarrow h(t) = 0$ , for  $t < 0$ . (Right hand side signal). (RHS)

← إذا كانت قيمته لـ  $(t < 0)$  يسو



NOTE:- Causal signal  
تحتل عن  
Causal system.

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau.$$

From the claim ( $h(t)$  is causal signal,  $h(t) = 0, t < 0$ ).

$$h(t-\tau) = \begin{cases} 0, & t-\tau < 0 \\ \text{value}, & t-\tau > 0 \end{cases} = \begin{cases} 0, & t < \tau \\ \text{value}, & t > \tau \end{cases}$$

$$y(t) = \int_{-\infty}^t x(\tau) h(t-\tau) d\tau.$$

The system is causal iff  $h(t)$  is a causal signal.

#### 4) Stable

$$|x(t)| < M \rightarrow |y(t)| < K$$

$y(t)$  will be bounded iff  $\int_{-\infty}^{\infty} |y(\tau)| d\tau < \infty$ .

absolutely integrable Impulse response.

#### Special case

LTI system + stable and causal.

$$= \int_{-\infty}^0 |h(\tau)| d\tau + \int_0^{\infty} |y(\tau)| d\tau.$$

we only check this.

Ex.  $h(t) = e^{-3t} u(t)$



① is it causal system?

yes, since  $h(t)$  is (RHS) signal.

② Stability -  $h(t)$  should be abs. integrable.

$$\int_{-\infty}^{\infty} |h(t)| dt < \infty \Rightarrow \int_0^{\infty} |e^{-3t} u(t)| dt = \left. -\frac{1}{3} e^{-3t} \right|_0^{\infty} = -\frac{1}{3} (0 - 1) = \frac{1}{3}$$

So yes it's stable.



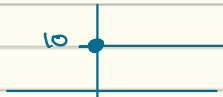
EX.  $h(t) = e^{3t} u(t)$



1 Causality :- it's Causal since its R.H.S. signal. ( $h(t) = 0$  for  $t < 0$ ).

2 Stability :-  $\int_0^{\infty} e^{3t} dt = \frac{1}{3} e^{3t} \Big|_0^{\infty} = \infty$ , so it's not stable.

EX.  $h(t) = 10 u(t)$



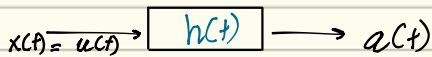
1 Causality :- yes, its causal since  $h(t) = 0$  for  $t < 0$ .

2 Stability :-  $\int_0^{\infty} 10 dt = 10(\infty) = \infty$ , so it's not bounded then it's not stable.

lecture 16 22:05 Dr. Qadri's Lectures

**Unit step response**

step response  $\rightarrow a(t)$



$$a(t) = u(t) * h(t)$$

$$= \int_{-\infty}^{\infty} u(\tau) h(t-\tau) d\tau$$

$$= \int_0^{\infty} h(t-\tau) d\tau$$

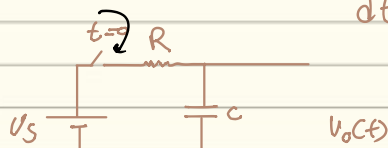
suppose that the system to be causal  $\Rightarrow$  so  $h(t)$  is Causal system  $= 0, t < 0$ .

Hence,  $h(t-\tau) = 0, t < \tau$

$a(t) = \int_0^t h(\tau) d\tau$   $\leftarrow$  if the signal is causal, I can use this integration instead

$\rightarrow$  the unit step response can be calculated directly from the unit impulse response  $h(t)$ .

\* we can show that  $h(t) = \frac{d a(t)}{dt}$



$v(t) = v_s u(t)$   
 step response  $v_o(t)$   
 $h(t) = \frac{d v_o(t)}{dt}$



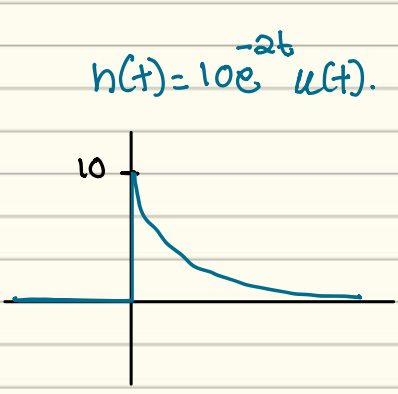
\*  $y(t) = a(t) * \frac{dx(t)}{dt}$

Ex. let  $h(t) = 10e^{-2t} u(t)$ , Find  $a(t)$ .

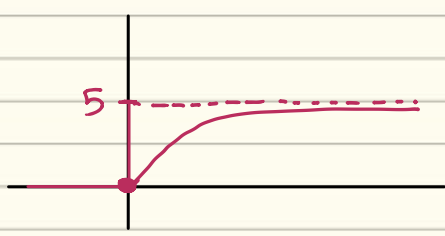
1. is it Causal system. yes, since  $h(t) = 0, t < 0$ .

2.  $a(t)$ .

$$a(t) = \int_0^t h(\tau) d\tau = \int_0^t 10e^{-2\tau} d\tau = \frac{10}{-2} e^{-2\tau} \Big|_0^t = -5(e^{-2t} - 1) = 5 - e^{-2t}$$



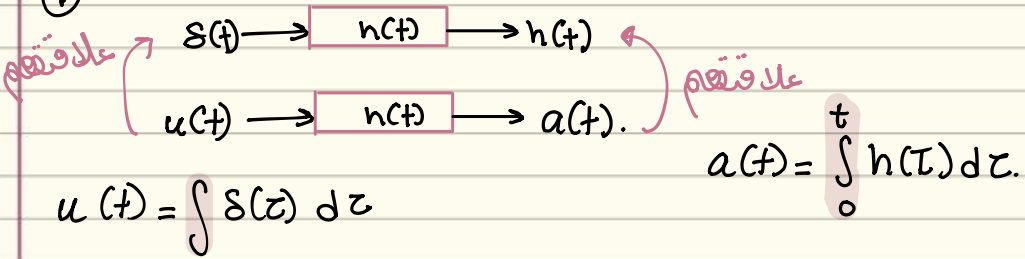
$a(t)$  step response.



Summary 8-

1. For LTI system, any linear operation on the input, produces the same linear operation on the output.

①



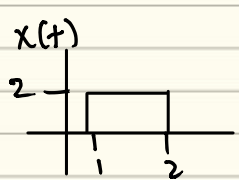
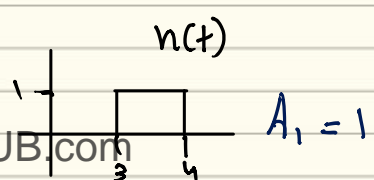
$$u(t) = \int \delta(\tau) d\tau$$

$$② \quad r(t) = \int u(t) dt \longrightarrow \int a(t)$$

⇒ if  $A_1$  is the Area under  $n(t)$ .  
 $A_2$  is the Area under  $x(t)$ .

بنقدر نستخرجها  
 عشان نتحقق  
 من الحل.

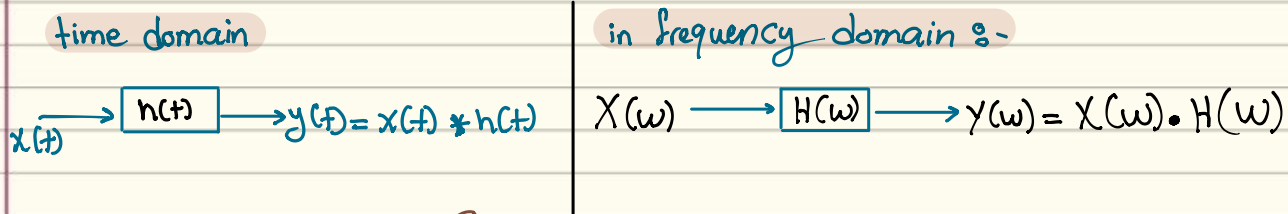
then, the area under  $y(t) = n(t) * x(t) = A_1 \cdot A_2$



$$A_y = A_1 \cdot A_2 = 2$$



# frequency response of LTI system



$X(t) \rightarrow X(\omega)$   
 $h(t) \rightarrow H(\omega)$

} on ch. 3 + ch. 4.

to prove that :-

$e^{j\omega t} \rightarrow \boxed{\phantom{H(\omega)}} \rightarrow y(t) = h(t) * e^{j\omega t}$

*Rotating phasor*

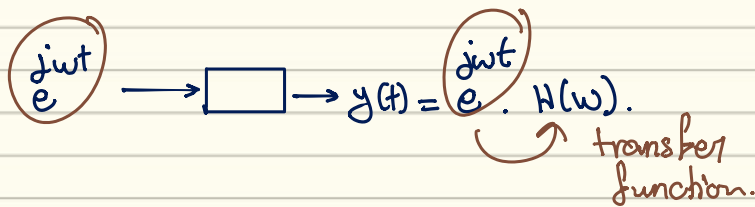
$$= \int_{-\infty}^{\infty} e^{j\omega(t-\tau)} \cdot h(\tau) d\tau$$

$$= \int_{-\infty}^{\infty} e^{j\omega t} \cdot e^{-j\omega\tau} h(\tau) d\tau$$

$$= e^{j\omega t} \cdot \int_{-\infty}^{\infty} e^{-j\omega\tau} h(\tau) d\tau$$

$$= e^{j\omega t} \cdot H(\omega)$$

$H(\omega) = \int_{-\infty}^{\infty} h(\tau) e^{-j\omega\tau} d\tau \rightarrow$  Frequency Response of the system  
 "Transfer function."



$H(\omega) = \text{Complex Quantity} \equiv \underbrace{|H(\omega)|}_{\text{Amplitude}} e^{j\Theta_H(\omega)}$

*phase*

*Frequency Response.*

## Summary :-

- if the input to a LTI system is a sinusoid ( $e^{j\omega t} = \cos(\omega t) + j\sin(\omega t)$ ), if frequency  $\omega$  rad/s, the steady state response is a sinusoid of the same frequency with amplitude mult. by  $|H(\omega)|$  a phase shifted by  $\Theta_H(\omega)$  radian.



من للفهدر  
الأنف

$|H(\omega)| = \frac{1}{\sqrt{1 + (\omega RC)^2}}$

$\Theta_{H(\omega)} = -\tan^{-1}(\omega RC)$

Example 2 - Find the response of  $x(t) = \underbrace{\cos(2000\pi t)}_{(1)} + \underbrace{\cos(4000\pi t)}_{(2)}$ .

$$\text{let } \frac{1}{2\pi RC} = 1 \text{ kHz}, \quad RC = 1000, \quad A = 1$$

$$y_1(t) = A |H(\omega)| \cdot \cos(2000\pi t + \theta)$$

$$RC = \frac{1}{2\pi(1000)} = \frac{1}{2000\pi}$$

$$|H(\omega)| = \frac{1}{\sqrt{1 + (\omega RC)^2}} = \frac{1}{\sqrt{1 + \left(\frac{2000\pi \cdot 1}{2000\pi}\right)^2}} = \frac{1}{\sqrt{2}} = 0.707$$

$$\theta = -\tan^{-1}\left(\frac{2000\pi \cdot 1}{2000\pi}\right) = -\pi/4.$$

$$H(\omega) = 0.707 e^{-j\pi/4}$$

$$\therefore y_1(t) = 0.707 \cos(2000\pi t - \pi/4).$$

doing the same for  $y_2$ .

$$|H(4000\pi)| = \frac{1}{\sqrt{1 + \left(\frac{4000\pi \cdot 1}{2000\pi}\right)^2}} = 0.45$$

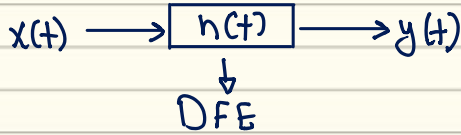
$$\theta(4000\pi) = -\tan^{-1}\left(\frac{4000\pi}{2000\pi}\right) = -63.43$$

$$y(t) = 0.707 \cos(2000\pi t - 45) + 0.45 \cos(4000\pi t - 63.45).$$



# simulate modeling

كتابة ال (system) عن طريق blocks.



$$\sum_{i=0}^n a_i \frac{dy}{dt^n} = \sum_{i=0}^n b_i \frac{dx(t)}{dt^n}$$

\* يفضل أعمل (integration) لحديته  
 مأوكل الفناش بكتابة y(t)  
 يعني الأوردر تليس = 0.

Example 8- sketch the simulating model for the following differential equation 8-

1)  $-x(t) + \frac{L}{R} \frac{dy(t)}{dt} + y(t) = 0$

Ans 8-  $-x(t) + \frac{L}{R} \frac{dy(t)}{dt} + y(t) = 0$ .

① بخلي ال [higher order] طرف والناقي على طرف ثاني.

بنحطها لخالصا.

بنجعل المعامل = 1  $\left( \frac{L}{R} \frac{dy(t)}{dt} = x(t) - y(t) \right)$

② أجعل معالمة = 1.

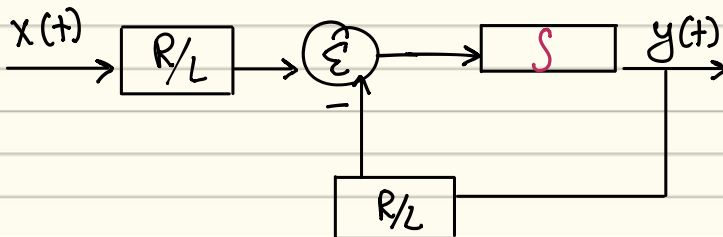
$$\frac{dy(t)}{dt} = \frac{R}{L} x(t) - \frac{R}{L} y(t)$$

③ نفترض ال الأشياء الي ما فيها مشتقة كالها متغير جديد  $q_0, q_1, \dots$

نقال:

$$y(t) = \int q_0$$

④ نقال.



## Example 2 :-

$$2 \frac{d^3 y(t)}{dt^3} - 8 \frac{d^2 y(t)}{dt^2} + 4 \frac{dy(t)}{dt} + 2y(t) = 4 \frac{dx(t)}{dt} + 2x(t)$$

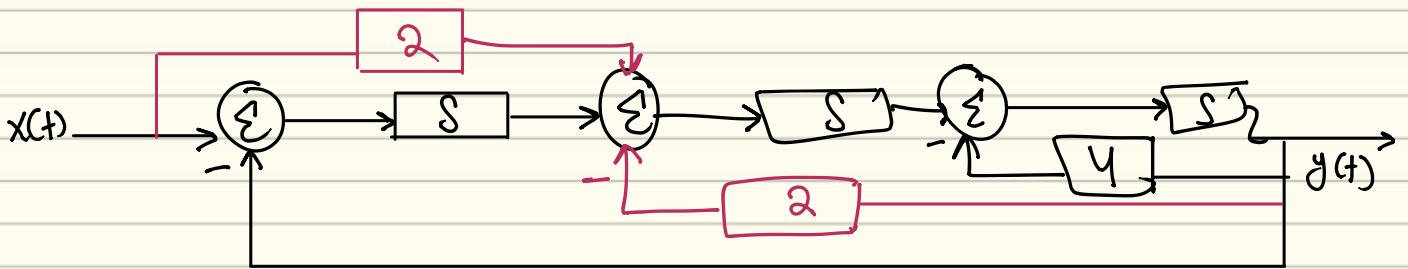
$$2 \frac{d^3 y(t)}{dt^3} = \frac{4 dx(t)}{dt} + 2x(t) + \frac{8 d^2 y(t)}{dt^2} - 4 \frac{dy(t)}{dt} - 2y(t)$$

$$\frac{d^3 y(t)}{dt^3} = \underbrace{x(t) - y(t)}_{q_0} + 2 \frac{dx(t)}{dt} - 2 \frac{dy(t)}{dt} + 4 \frac{d^2 y(t)}{dt^2}$$

$$\frac{d^2 y(t)}{dt^2} = \underbrace{\int q_0 + 2x(t) - 2y(t)}_{q_1} + 4 \frac{dy(t)}{dt}$$

$$\frac{dy(t)}{dt} = \underbrace{\int q_1 + 4y(t)}_{q_2}$$

$$y(t) = \int q_2$$



ربنا تقبل منا إنك أنت السميع العليم

روان فارس

