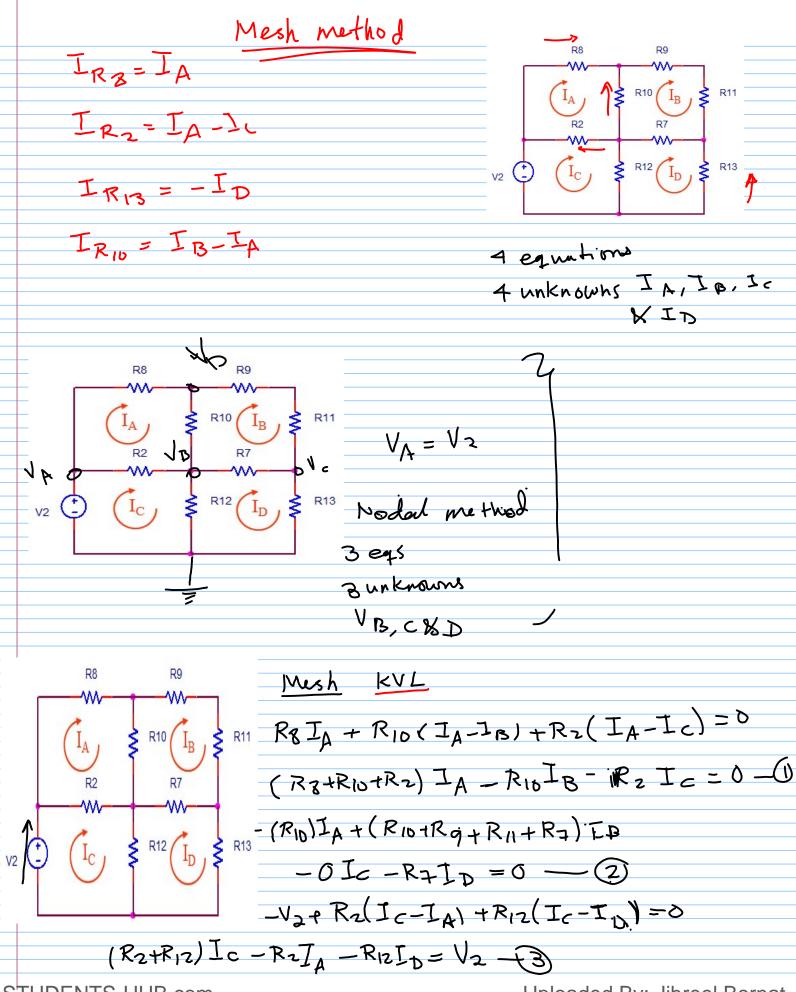
12-41512 Nodal method  $V_2$ @ node Vi \$ 10 m L 2Ω All currents are out 2 URIJ V lout  $3 + \frac{V_1}{2} + \frac{V_1 - V_2}{2} + 2 = 0$  $> I_{aut} = 0$ OR J 10 = 0  $(1.5) V_1 - V_2 = -5 - (1)$ @ Node V2  $0 \sqrt{1} - 2 + \frac{\sqrt{2} - \sqrt{1}}{1} + \frac{\sqrt{2}}{5} = 0$  $V_1 = -5$  $-V_{1+}$  ). 2  $V_2 = 2$  -2V2=-2.51 @ node VI  $\left(\frac{1}{R_1}, \frac{1}{R_2}\right) V_1 = \left(\frac{1}{R_1}\right) V_2 = T_{S_1}$ @ node V2  $-\left(\frac{1}{R_1}\right)V_1+\left(\frac{1}{R_1}+\frac{1}{R_3}\right)V_2=I_{52}$ 

@ node 1  $\Sigma I out = \Sigma I in$  $7V_1 - 3V_2 - 4V_3 = -11 - ()$ @ node V2  $3(v_1-v_2) + 4(v_1-v_3) + 3 + 8 = 0$  $-(3)V_1 + (3+2+1)V_2 - (2)V_3 = 3 - 2$ anode V3  $-4V_{1}-2V_{2}+(5+2+4)V_{3}=25$ 6r V3 2-2 V١  $\xi N_{\mathcal{N}}$ 32 2A 7)100 I out = 2 cm 205  $+\frac{V_{1}}{10}+\frac{V_{1}-V_{2}}{10}=$ KUL -V2 + 2 IX + 10 = 0 Z Iout=0  $-1_{x} = \frac{V_{2} - 10}{2}$  $\frac{V_{2} - V_{1}}{4} + \frac{V_{2}}{3} + \frac{V_{2} - 10}{2} = 0$ V, = -

Special Gase 32V1 100 V2 6V EX 210 54\_r @node Vi 201 4  $-\lambda + \frac{v_1}{5} + I_X = 0$  $-2 + \frac{V_1}{5} + \frac{V_2}{10} + \frac{V_2 + 6}{4} = 0 - \frac{1}{2}$  $|\langle v \rangle|$ (2) KVL (a supernode  $V_{1} = V_{2} = 10$ G-X Ι  $\int_{-\infty}^{\infty} = \frac{\sqrt{2}}{3k} = 0$ **З**12К Q rode V,  $2 \text{ Io} + \frac{V_1}{12k} + \frac{V_1 - V^2}{6k} = 0$  $2(\frac{V_{2}}{3k}) + \frac{V_{1}}{Rk} + \frac{V_{1}-V_{2}}{6k} = 0$  (1) رتها @ node V2 Va-VI + V2 + 2mA=0 (2) y-1  $V_1 := -\frac{24}{5}V \quad V_1 := \frac{12}{5}V$ 



(RIL+R7+RI3)ID-0IA-R7IB-RIZIC=0- 4)  $-42 + 6I_{1} + 3(I_{1} - I_{2}) = 0$ Figure3 example 1 of mesh analys  $9I_1 - 3I_2 = 42 - (1)$ 3I, +7I2 = 10 - 3 6A, 4A special case 1 current source exists only in one mesh 40 ΕX  $T_2 = -5A_{\nu}$ 10 10 + 4I1+6(I,=5)=0 Figur6:mesh with current sourcee J, = -2A STUDENTS-HUB.com Uploaded By: Jibreel Bornat

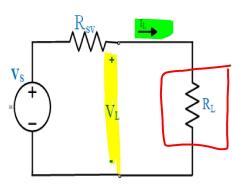
## Case 2:

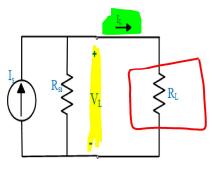
Current source exists between two meshes, a Super mesh is obtained pKVL for mesh(2)  $(1+2+3)I_2 - (1)I_1 - (3)I_3 = 0 -$ 1° constraint equation I<sub>3</sub> I, - I3 = 7 - 2 > Super mesh equation I -- - 7A X  $-7 + (1)(I_1 - I_2) + (3)(I_3 - I_2) + (1)I_3 = 0$  $I_1 - 4 I_2 + 4 I_3 = 7 - (3)$ GR -7+212+1,=0-**EX** Mesh Analysis with dependent sources  $\sqrt{x} = (3)(I_3 - I_2)$  $I_1$  $I_1 = (5A) V$  $T_2 + 2I_3 = 45$ p constraint equation  $\frac{V}{q} = I_3 - I_1$  $\frac{3}{9}(I_3 - I_2) = I_3 - 15 | (1)(I_2 - 15) + (2)I_2 + (3) (I_2 - I_3) = 0$  $6I_{2} - 3I_{3} = 15$   $-2I_{2} = 11AV$  $I_{3} = 17AV$  $\frac{3}{2}I_2 + \frac{2}{3}I_3 = 15$ 

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- Node or mesh: How to choose?
- $\triangleright$  Use the one with fewer equations.
- $\triangleright$  Use the method you like best.

Source Transformation





slides

Two sources are equivalent, if each produces identical current and identical voltage in any load which is placed across its terminal.

plat R\_= Zero (short circuit s.C.) Isc, Rev Iscz Vz ٢ 3 RFI V, If ٧. Rsi Iscz=  $\Gamma_{SCI} = \frac{V_{S}}{R_{SV}}$ ISC2 = Is V,= V2 = Zero Isci = Isca For Ir= Vs Rr  $\oplus$ STUDENTS-HI Uploaded By: Jibreel Bornat

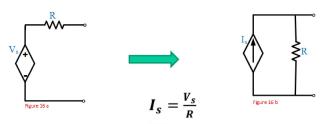
1,= 12 = 2000 plet RL= 00 (open circuit) **I**2 RFV Voci 75 ٧ş Rsi VO.C.2 Voci = V q Vocz = Is Rsi For Voci = Vocz Vr = IsRsi Vs Rsi Is = Ir = Vs GR Ð Rsi = REV 0 28 ξ5 44 ک 510 find Vo ΕX\ 2ra Sr t)20V 10 V V, ₽ 9A \$ 10 p 4~ 55 04 -.9)(4) = Vo 음 오 322 4\_2 V=9.47 Volt. 5A h

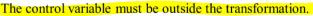
25  $\langle \neg \rangle$ a OR + 1 10 V 20 V Noda V<u>0 - 7</u>8 Vo - X  $\leq 0$ V<u>0 - 10</u> Y 0 V, ( + -+ 0.95  $\bigvee$ Vo = Volt 4-3 Therenin X •b Norton eq circuit (a) • a ŧ ₹R ₹R (b) Figure: 04-39a,b Copyright © 2008 Pearson Prentice Hall, Inc.

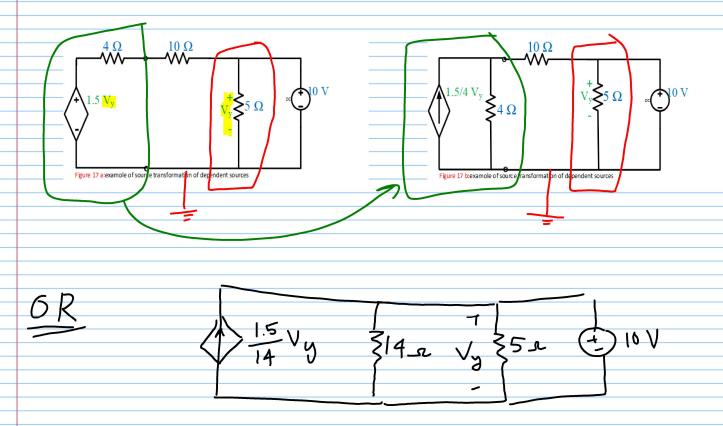
- The two circuits depicted in Fig. 4.39(a) are equivalent with respect to terminals a,b because they produce the same voltage and current in any resistor *RL* inserted between nodes a,b.
- The same can be said for the circuits in Fig. 4.39(b).

 $25 \Omega$  $5\,\Omega$ Example: Find Vo using source transformation w neglecte 8 A 250 V ່ **v**<sub>o</sub> ຊ\$100 Ω \$15 Ωີ 125 ( ₹10Ω Figure: 04-40Ex4.9 Copyright © 2008 Pearson Prentice Hall, Inc 25- DBA 3100 6 20 10 A 20-\$ 25 m √. \$ 10 m 2A Vs=(2A) (10 r) = 20 Volt. +  $\vee_{\bullet}$ 210-e 2A

Dependent sources







## The Superposition Theorem

In a linear network, the voltage across or the current through any element may be calculated by adding algebraically all the individual voltages or currents caused by the separate independent sources acting alone, i.e. with:

1) All other independent voltage sources replaced by short circuits.

2) All other independent current sources replaced by open circuits.

Dependent sources are left intact because they are controlled by circuit variables.

## Steps to apply superposition principle

- Turn off all independent sources except one source. Find the output (voltage or current) due to that source using nodal, mesh.....
- Repeat step 1 for each of the other independent sources.
- Find the total contribution by adding algebraically all contributions
  due to each independent sources.

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2

 $7.4 \mathrm{K}\,\Omega$ 3kΩ χı  $\sim$ find (I) using 9 m.A 5KΩ superposition Figure15 "a" due to (9mf) 5K 9m 3K-r 5K+7.41231 4.4Kr 2.922 mA 9mA 351< 2) Iz due to (30) - 3 3Kr  $I_2 =$ 7.4K-1 15.4K O mA = -0.19 12 3V 55K-A = 1 + 12U 00 = 2.992 - 0.194= 2.728 mA

Use superposition to solve for ix

100-0 6Ω  $i_x >_{9\Omega}$ 20 20 6 ~ gure1:"a":example of superposition ix, due to (3V) 30(+  $(\mathcal{I})$ ۍ ۹ چ l×1  $i_{x_1} = \frac{3}{9+6} = 0.2 \text{ A}$ 100-) Z A \$6 ~ \$ \$ 9 ~ @ 1x2 due to 2A IXZ  $i_{X_2} = \frac{6}{4+9} = 2A = 0.8A$ LX = 1X1+1X2 = 0.2+0.8 = 1A a 00 EXI 1 25 12 V 10 V 1 x ЗA find Ix using superposition. -یر کر Oix, due to (10 V) ly. 10 10 lx,

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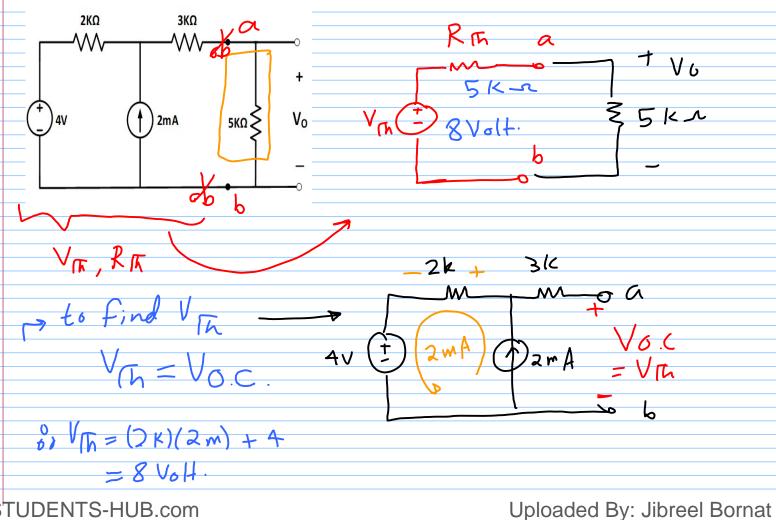
@ ixz due to (3A)  $\frac{2}{1}$   $\frac{1}{1}$   $\frac{1}{2}$   $\frac{1}$  $i_{X_2} = \frac{1}{2} - 3 = -1$ A 3 ix due to 12V , r JIZ V ix\_= -= - 4A & Ix = (x, + (x2 + 1 × 3 - 3.33-1-4 - - 1.666 A 1 EX\ 25 -> + 10 V Ιx + )3A 5 v find Ix using superposition. کر ۱ 2~~ () ix, due to (10V) 10 (± lxi  $i_{x_1} = \frac{10}{2} = 5A$ 

(2) i x2 due to (5v) (x=-==-2.5A ±)5v LX7. 3 (x3 due to (3A)  $\backslash \mathcal{N}$ 2 r 1X3= + ero ЗA  $ly_{z}$  $l_{X} = l_{X1} + l_{X2} + l_{X3}$ 6 Y = 5-2.5+0 - 2.5A  $2 \Omega$ Find *Ix* using superposition 2 Ix O ix, due to 105 Figure1"a":Superposition with a dependent source  $\backslash \mathcal{N}$  $2 \wedge$ KVL ZXIXI -10+3(x1+2(x1=0 ∼ 101 (+) LXI i x = 2A

1~ (2) ix2 due to (3A) Ξγ × 21×2 ) З A (XJ KVL 2 IX2 - IY +21x2=0 4 1x2 - Iy = 0 KCL IX2+Iy=3 5 Jyn = 3 1×2 33/5 =-0.6A 0° Ix = (x1+1x2 = 2-0.6 = 1.4 A  $2 \wedge$ FX\Find Ix ίx ;21;  $l_{x} = 10A$ )34 10A OIX, due to (10A) Azixi (x, = 10A l X1 IOA

(2) 1×2 fue to (3A) \_\_\_\_\_ \_\_\_\_ \_\_\_\_\_ 2 . L/2=0 21×2 € 3A Therenin X Norton eq. circuits

Example: Find Vo using thevenin's equivalent circuit

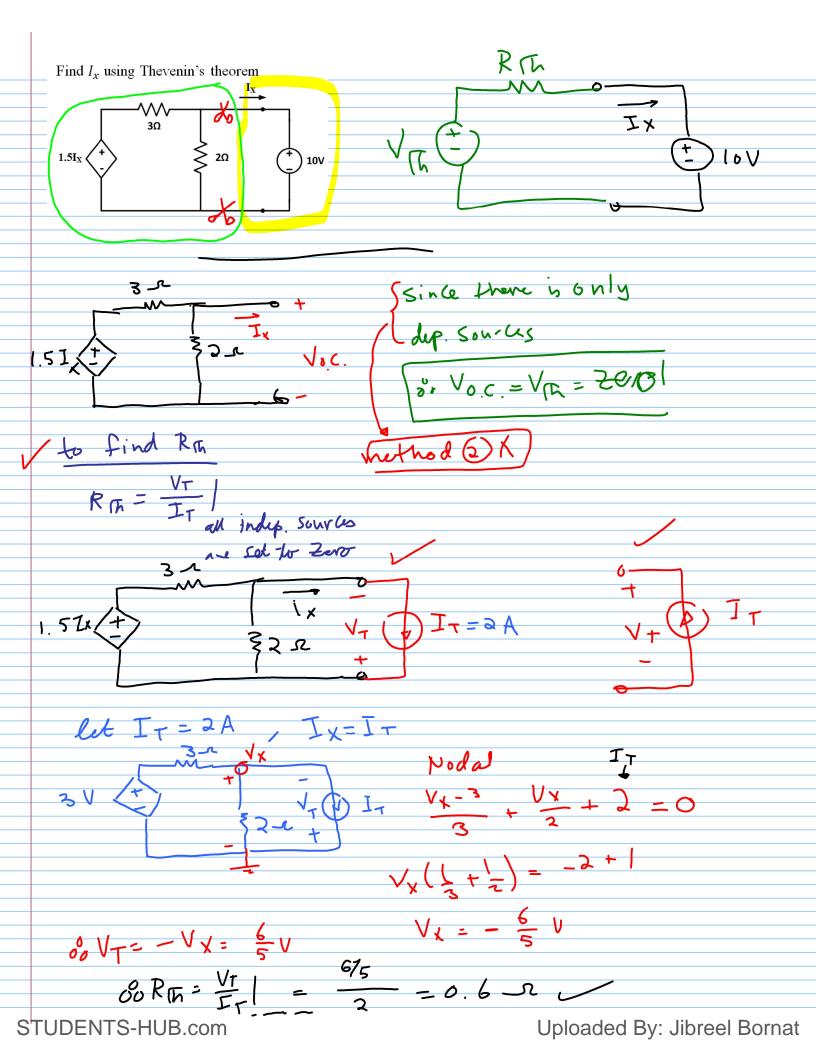


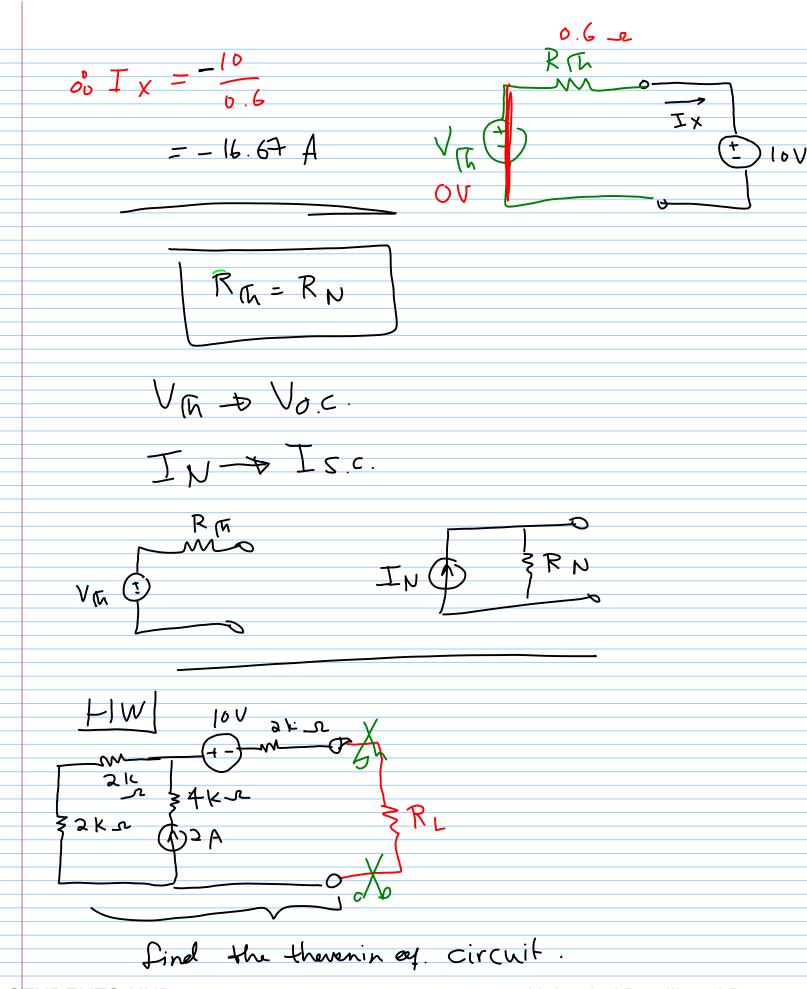
(No dependent sources) Rest method \* Kill all independent sources -> set to zero all indep. V sources S.C. (short circuit) allindup. I souvers A -> O.C. (open circuit) 3K ~ 2K r moa  $\frac{R_{\rm fh}}{1} = 5.2$ 1 + V6  $00 V_0 = \frac{5}{5+5} \times 8$ 35KJ 8Volt. All the sources this method when there is NO lep. sources, But - time consuminy!!) method (2) Vo.c. = 8V (the same as before)  $R_{fh} = \frac{V_{o.c.}}{I_{s.c.}}$ 

Find Is.c. SKV 3 R-L -0 Is.c. (= IN TIZMA (IS.C. 41 9 2KX(Is.c. - I, = 2mA)\_() -4 + 2K II + 3K Is.c. = 0  $v(3kIsc. +2kI_1 = 4) - (2)$ 5KJ10 = 8 BORTH = Vo.c. = 8 IS.C. = 8/2m Is.c. = % mA = 5K-2 = IN. method (3)  $R_{Th}$  using test sources  $R_{Th} = \frac{V_T}{I_T} |_{all index sources are sources}$ 2K r moa set to zero let VT= 5V then IT = Vr = 5 = ImA "RT = 5Kr

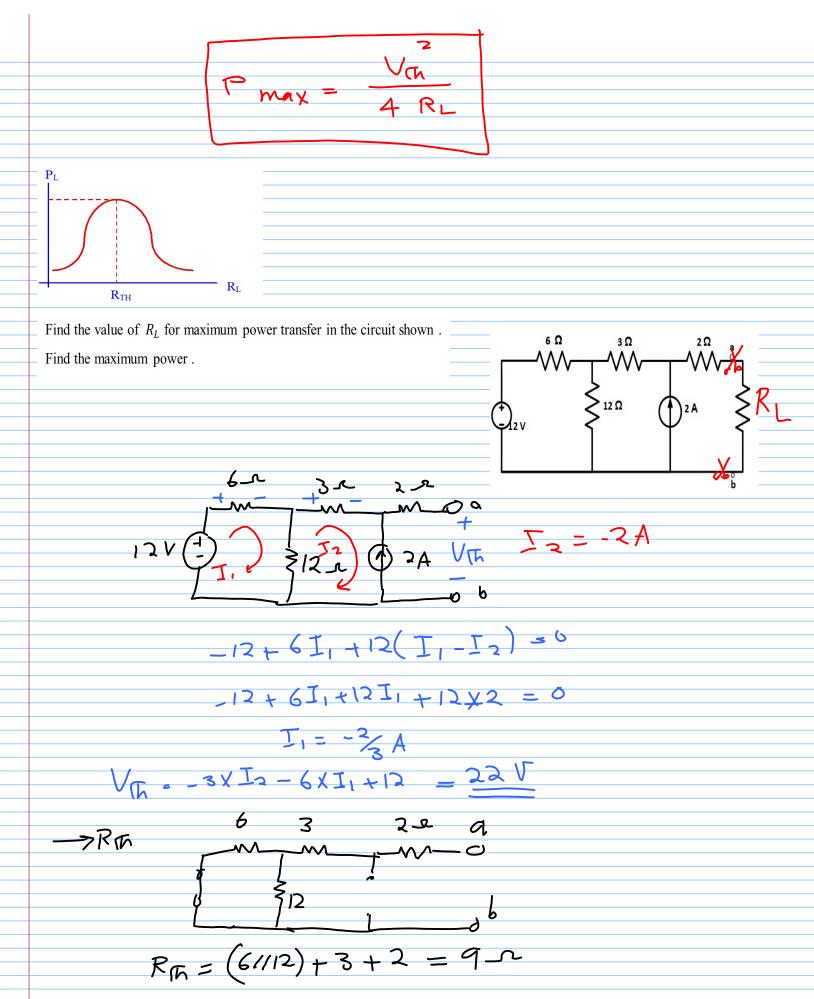
+ methods to Find RT. :-- Kill all indep sources, then find R. (series/parallel Combination) Smithod 2 RM = VO.C. Greneval method can be applied if the circuit has dep. sources (also it can be applied if the circuit does not have dep. sources)  $\frac{1}{T_{T}} = \frac{V_{T}}{T_{T}} \left( \frac{dep. sources one left}{dep. sources} \right)$ Find  $V_o$  using Thevenin's theorem 2K 3K a For this Circuit  $V_{\rm X}$ 4000 V  $V_{\rm X}$   $V_{\rm L} = V_{\rm O.C.}$  $V_{\text{fh}} = V_{0.C.} = V_{X}$ -**o**-0,  $\frac{V_{o,c}}{(4\pi)} = \left(\frac{V_{x}}{(4\pi)}\right) 2k + 4V = \left(\frac{V_{o,c}}{(4\pi)}\right) 2k + 4^{1}$  $m = \frac{1}{2} V_{0,c+4} = \frac{V_{0,c-3} V_{0,c-1}}{V_{0,c-3} V_{0,c-3}}$ STUDENTS-HUB.com Uploaded By: Jibreel Bornat

to find RTh \_ method 1 X method 2 or method 3 V (dep. source) method @ Vo.C. = 8V. Ú 3K 2 K find Isc.  $\mathbf{M}$ M\_ VX = Zero!! VX 4000 VX 4V  $\frac{V_x}{4060} = 0 A(0.c.)$ -0ú ZK 3K 8. Is.c. = 4 = 0.8 mA 4 V 00 R/h = VO.C. = 8 IS.L. = 0.3m = 10 K -2 method 3  $R_{Th} = \frac{V_T}{I_T}$ all indep. sonvers 12.11ed Ú 3k 2Klet VT = 4000 V : I- = 4000 - 2000 TVE VX 4000 VX IT = 0.4 A  $\hat{v} R = \frac{4000}{0.4} = 10 \text{ K} \text{ K}$   $= \frac{4000}{0.4} = 10 \text{ K} \text{ K}$ -0-01 JK-2 3K e-M T7 F) VT (±)VT 6 2000 4000 1 A 4000 V





4.12 Maximum Power Transfer Rr  $\Lambda \Lambda \Lambda$ Vr C circuit RL max Power - RL  $P = i R L = \left(\frac{V_R}{D_1 R}\right) R L$ Pipmax when  $\frac{dP}{dR_{L}} = Zero$   $\frac{dR_{L}}{R_{L}} = \frac{V_{R}}{R_{L}}$ (RL+RM)  $\frac{dP}{dR_L} = V_{fh} \left[ \frac{(R_{fh} + R_L) - R_L \chi_2 \chi(R_L + R_{fh})}{(R_{fh} + R_L)^{\dagger}} \right] = 0$  $\left(\mathcal{R}_{\mathrm{IR}} + \mathcal{R}_{\mathrm{L}}\right) = 2 R_{\mathrm{L}} \left(\mathcal{R}_{\mathrm{L}} + \mathcal{R}_{\mathrm{R}}\right)$  $R_{t}^{2} + R_{1}^{2} + 2R_{2}R_{m} = 2R_{1}^{2} + 2R_{1}R_{m}$ RL=RT - The wax power transfer occurs when the load resistance RL = Rth or Pmax= (VA) RL



RK=9-52 Vĥ SR1 6 225 of for mak. Power transfer RL = RE = 95  $\frac{0}{00} P_{mad} = \frac{V_{R}^{2}}{4R_{R}} = 13.44W$ 91 if RL= 6-2 22V RL  $P_{RL} = C RL$  $=\left(\frac{22}{6+6}\right) \times 6$ if RL= 13-2 then  $P_{RL} = \left(\frac{22}{9+13}\right) \times 13$ = 12.9W 9 = 13W >RL 9 r