

Problem statement

Problem: Given a nonlinear system $\dot{X} = f(X, U)$

Derive an approximate linear system $\dot{X} = AX + BU$

about an "Operating Point" (X_0, U_0)

Note: An operating point is a point through which the system trajectory passes.

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Linearization: Scalar homogeneous systems

 $\dot{x} = f(x), \quad x \in R$ Scalar system: **Operating point:** x_0 Define: $x = x_0 + \Delta x$ **Taylor series:** $\dot{x}_{0} + \Delta \dot{x} = f(x_{0} + \Delta x) = f(x_{0}) + f'(x) \Big|_{x_{0}} \Delta x + \left\{ f''(x) \frac{(\Delta x)^{2}}{2!} + \cdots \right\}$

Neglecting HOT,
$$\dot{x}_0 + \Delta \dot{x} \approx f(x_0) + f'(x_0)\Delta x$$

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

3

HOT

Linearization: Scalar homogeneous systems

 x_0 satisfies the differential equation $\dot{x}_0 = f(x_0)$ This leads to $\Delta \dot{x} = [f'(x_0)]\Delta x = a \Delta x$

For convenience, redefine $x \triangleq \Delta x$

This leads to

$$\dot{x} = ax$$

where $a = f'(x)$

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

Example – 1

Linearize:

Solution:

$$\dot{x} = x^2 - 1, \quad x(0) = \pm 1$$

 $a_1 = \frac{df}{dx}\Big|_{x_0 = 1} = 2x_0 \Big|_{x_0 = 1} = 2$

$$a_2 = \frac{df}{dx} |_{x_0 = -1} = 2x_0 |_{x_0 = -1} = -2$$

The linearized system:

 $\dot{x} = 2x$ $x_0 = 1$

$$\dot{x} = -2x \qquad x_0 = -1$$

Note: As the reference point changes, the linearized approximation also changes!

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Linearization: General homogeneous systems

Homogeneous System:

$$\dot{X} = f(X), \quad f \triangleq \begin{bmatrix} f_1 & f_2 & \dots & f_n \end{bmatrix}^T, \quad X \triangleq \begin{bmatrix} x_1 & x_2 & \dots & x_n \end{bmatrix}^T$$
Taylor Series:
$$f(X_0 + \Delta X) = f(X_0) + \begin{bmatrix} \frac{\partial f}{\partial X} \end{bmatrix}_{X_0} \Delta X + HOT$$

$$\dot{X}_0 + \Delta \dot{X} \approx f(X_0) + \begin{bmatrix} \frac{\partial f}{\partial X} \end{bmatrix}_{X_0} \Delta X$$

$$\Delta X \triangleq X \qquad \qquad A = \begin{bmatrix} \frac{\partial f}{\partial X} \end{bmatrix}_{X_0} \triangleq \begin{bmatrix} \frac{\partial f_1}{\partial x_1} & \dots & \frac{\partial f_1}{\partial x_n} \\ \vdots & \ddots & \vdots \\ \frac{\partial f_n}{\partial x_1} & \dots & \frac{\partial f_n}{\partial x_n} \end{bmatrix}$$

ADVANCED CONTROL SYSTEM DESIGN

Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

Example – 2: Van-der Pol's Oscillator (Limit cycle behaviour)

- Equation $M \ddot{x} + 2c(x^2 1)\dot{x} + k x = 0$ $\{c, k > 0\}$
- State variables $x_1 \triangleq x$, $x_2 \triangleq \dot{x}$
- State Space Equation

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} x_2 \\ -\frac{2c}{m} \left(x_1^2 - 1 \right) x_2 - \frac{k}{m} x_1 \end{bmatrix}_{F(X)}$$

: Homogeneous nonlinear system

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

Example – 2: Van-der Pol's Oscillator (Limit cycle behaviour)

• Operating Point: $\int \chi_{1_0}$

$$x_{1_0} \quad x_{2_0} \end{bmatrix}^T = \begin{bmatrix} 0 & 0 \end{bmatrix}$$

Linearized State Space Equation



ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

Linearization: General Systems

System having control input

 $\dot{X} = f(X,U), \qquad f, X \in \mathbb{R}^n, \quad U \in \mathbb{R}^m$

Reference point: (X_0, U_0) Taylor series expansion:

$$f\left(X_{0} + \Delta X, U_{0} + \Delta U\right)$$

= $f\left(X_{0}, U_{0}\right) + \left[\frac{\partial f}{\partial X}\right]_{(X_{0}, U_{0})} \Delta X + \left[\frac{\partial f}{\partial U}\right]_{(X_{0}, U_{0})} \Delta U + HOT$

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

Linearization

$$\dot{X}_{0} + \Delta \dot{X} \approx f(X_{0}, U_{0}) + \left[\frac{\partial f}{\partial X}\right]_{(X_{0}, U_{0})} \Delta X + \left[\frac{\partial f}{\partial U}\right]_{(X_{0}, U_{0})} \Delta U$$
$$\Delta \dot{X} = A \Delta X + B \Delta U$$
Re-define: $\Delta X \triangleq X, \quad \Delta U \triangleq U$

This leads to X = AX + BU



ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

10

STUDENTS-HUB.com

Example - 3: Spinning Body Dynamics (Satellite dynamics)

Dynamics:

$$\dot{\omega}_{1} = \left(\frac{I_{2} - I_{3}}{I_{1}}\right) \omega_{2} \omega_{3} + \left(\frac{1}{I_{1}}\right) \tau_{1}$$
$$\dot{\omega}_{2} = \left(\frac{I_{3} - I_{1}}{I_{2}}\right) \omega_{3} \omega_{1} + \left(\frac{1}{I_{2}}\right) \tau_{2}$$
$$\dot{\omega}_{3} = \left(\frac{I_{1} - I_{2}}{I_{3}}\right) \omega_{1} \omega_{2} + \left(\frac{1}{I_{3}}\right) \tau_{3}$$

 I_1, I_2, I_3 : MI about principal axes $\omega_1, \omega_2, \omega_3$: Angular velocities about principal axes

 τ_1, τ_2, τ_3 : Torques about principal axes

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

Example - 3: Spinning Body Dynamics (Satellite dynamics)

• Operating Point:
$$\begin{bmatrix} \omega_{1_0} & \omega_{2_0} & \omega_{3_0} \end{bmatrix}^T = \begin{bmatrix} 0 & 0 & 0 \end{bmatrix}^T$$

 $\begin{bmatrix} \tau_{1_0} & \tau_{2_0} & \tau_{3_0} \end{bmatrix}^T = \begin{bmatrix} 0 & 0 & 0 \end{bmatrix}^T$

Linearized State Space Equation (Double Integrator) •

$$\begin{bmatrix} \dot{\omega}_{1} \\ \dot{\omega}_{2} \\ \dot{\omega}_{3} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \\ \dot{\omega}_{3} \end{bmatrix} \begin{bmatrix} \omega_{1} \\ \omega_{2} \\ \omega_{3} \end{bmatrix} + \begin{bmatrix} (1/I_{1}) & 0 & 0 \\ 0 & (1/I_{2}) & 0 \\ 0 & 0 & (1/I_{3}) \end{bmatrix} \begin{bmatrix} \tau_{1} \\ \tau_{2} \\ \tau_{3} \end{bmatrix}$$

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

Example – 4: Airplane Dynamics, Six Degree-of-Freedom Nonlinear Model

Ref: Roskam J., Airplane Flight Dynamics and Automatic Controls, 1995

$$\begin{split} \dot{U} &= VR - WQ - g\sin\Theta + \left(F_{A_{X}} + F_{T_{X}}\right)/m \\ \dot{V} &= WP - UR + g\sin\Phi\cos\Theta + \left(F_{A_{Y}} + F_{T_{Y}}\right)/m \\ \dot{W} &= UQ - VP + g\cos\Phi\cos\Theta + \left(F_{A_{Y}} + F_{T_{Y}}\right)/m \\ \dot{P} &= c_{1}QR + c_{2}PQ + c_{3}\left(L_{A} + L_{T}\right) + c_{4}\left(N_{A} + N_{T}\right) \\ \dot{Q} &= c_{5}PR - c_{6}\left(P^{2} - R^{2}\right) + c_{7}\left(M_{A} + M_{T}\right) \\ \dot{R} &= c_{8}PQ - c_{2}QR + c_{4}\left(L_{A} + L_{T}\right) + c_{9}\left(N_{A} + N_{T}\right) \\ \dot{\Phi} &= P + Q\sin\Phi\tan\Theta + R\cos\Phi\tan\Theta \\ \dot{\Theta} &= Q\cos\Phi - R\sin\Phi \\ \dot{\Psi} &= \left(Q\sin\Phi + R\cos\Phi\right)\sec\Theta \\ \begin{bmatrix} \dot{X}' \\ \dot{Y}' \\ \dot{Z}' \end{bmatrix} = \begin{bmatrix} \cos\Psi & -\sin\Psi & 0 \\ \sin\Psi & \cos\Psi & 0 \\ 0 & 1 & 0 \\ -\sin\Theta & 0 & \cos\Theta \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 \\ 0 & \cos\Phi & -\sin\Phi \\ 0 & \sin\Phi & \cos\Phi \end{bmatrix} \begin{bmatrix} U \\ V \\ W \end{bmatrix} \quad \begin{bmatrix} Note: & \dot{h} = -\dot{Z}' \end{bmatrix}$$

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Linearization using Small Perturbation Theory

Perturbation in the variables: $U = U_0 + \Delta U$ $V = V_0 + \Delta V$ $W = W_0 + \Delta W$ $P = P_0 + \Delta P$ $Q = Q_0 + \Delta Q$ $R = R_0 + \Delta R$ $X = X_0 + \Delta X$ $Y = Y_0 + \Delta Y$ $Z = Z_0 + \Delta Z$ $X_T = X_{T_0} + \Delta X_T \quad Y_T = Y_{T_0} + \Delta Y_T \quad Z_T = Z_{T_0} + \Delta Z$ $M = M_0 + \Delta M$ $N = N_0 + \Delta N$ $L = L_0 + \Delta L$ $\Phi = \Phi_0 + \Delta \phi \qquad \Theta = \Theta_0 + \Delta \theta \qquad \Psi = \Psi_0 + \Delta \psi$ $\delta_A = \delta_{A_0} + \Delta \delta_A \quad \delta_E = \delta_{E_0} + \Delta \delta_E \quad \delta_R = \delta_{R_0} + \Delta \delta_R$

> ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Trim Condition for Straight and Level Flight

• Assume:
$$V_0 = P_0 = Q_0 = R_0 = \Phi_0 = Y_{T_0} = Z_{T_0} = 0$$

- Select: X_{T_0}, z_{I_0} (*i.e.* h_0)
- Enforce: $\dot{U} = \dot{V} = \dot{W} = \dot{P} = \dot{Q} = \dot{R} = \dot{\Phi} = \dot{\Theta} = \dot{z}_I = 0$
- Solve for: $U_0, W_0, X_0, Y_0, Z_0, L_0, M_0, N_0, \Theta_0$

• Verify:
$$Y_0 = L_0 = M_0 = N_0 = 0$$

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

Uploaded By: Mohammad Awawdeh

Typically True $\forall t$

Linearization using Small Perturbation Theory

Reference: R. C. Nelson, Flight Stability and Automatic Control, McGraw-Hill, 1989.

$$\begin{split} \Delta X &= \frac{\partial X}{\partial U} \Delta U + \frac{\partial X}{\partial W} \Delta W + \frac{\partial X}{\partial \delta_E} \Delta \delta_E + \frac{\partial X}{\partial \delta_T} \Delta \delta_T \\ \Delta Y &= \frac{\partial Y}{\partial V} \Delta V + \frac{\partial Y}{\partial P} \Delta P + \frac{\partial Y}{\partial R} \Delta R + \frac{\partial Y}{\partial \delta_R} \Delta \delta_R \\ \Delta Z &= \frac{\partial Z}{\partial U} \Delta U + \frac{\partial Z}{\partial W} \Delta W + \frac{\partial Z}{\partial \dot{W}} \Delta \dot{W} + \frac{\partial Z}{\partial Q} \Delta Q + \frac{\partial Z}{\partial \delta_E} \Delta \delta_E + \frac{\partial Z}{\partial \delta_T} \Delta \delta_T \\ \Delta L &= \frac{\partial L}{\partial V} \Delta V + \frac{\partial L}{\partial P} \Delta P + \frac{\partial L}{\partial R} \Delta R + \frac{\partial L}{\partial \delta_R} \Delta \delta_R + \frac{\partial L}{\partial \delta_A} \Delta \delta_A \\ \Delta M &= \frac{\partial M}{\partial U} \Delta U + \frac{\partial M}{\partial W} \Delta W + \frac{\partial M}{\partial \dot{W}} \Delta \dot{W} + \frac{\partial M}{\partial Q} \Delta Q + \frac{\partial M}{\partial \delta_E} \Delta \delta_E + \frac{\partial M}{\partial \delta_T} \Delta \delta_T \\ \Delta N &= \frac{\partial N}{\partial V} \Delta V + \frac{\partial N}{\partial P} \Delta P + \frac{\partial N}{\partial R} \Delta R + \frac{\partial N}{\partial \delta_R} \Delta \delta_R + \frac{\partial N}{\partial \delta_A} \Delta \delta_A \end{split}$$

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

16

STUDENTS-HUB.com

State Variable Representation of Longitudinal Dynamics

Reference: R. C. Nelson, Flight Stability and Automatic Control, McGraw-Hill, 1989.

State space form:

 $\dot{X} = AX + BU_c$



Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

State Variable Representation of Lateral Dynamics

State space form: $X = AX + BU_{c}$ $A = \begin{bmatrix} Y_{V} & Y_{P} & -(U_{0} - Y_{R}) & g \cos \theta_{0} \\ L_{V}^{*} + \frac{I_{XZ}}{I_{X}} N_{V}^{*} & L_{P}^{*} + \frac{I_{XZ}}{I_{X}} N_{P}^{*} & L_{R}^{*} + \frac{I_{XZ}}{I_{X}} N_{R}^{*} & 0 \\ N_{V}^{*} + \frac{I_{XZ}}{I_{Z}} L_{V}^{*} & N_{P}^{*} + \frac{I_{XZ}}{I_{Z}} L_{P}^{*} & N_{R}^{*} + \frac{I_{XZ}}{I_{Z}} L_{R}^{*} & 0 \\ 0 & 1 & 0 & 1 \end{bmatrix} \qquad X = \begin{bmatrix} \Delta V \\ \Delta P \\ \Delta R \\ \Delta \phi \end{bmatrix}$ $B = \begin{bmatrix} 0 & Y_{\delta_{R}} \\ L_{\delta_{A}}^{*} + \frac{I_{XZ}}{I_{X}} N_{\delta_{A}} & L_{\delta_{R}}^{*} + \frac{I_{XZ}}{I_{X}} N_{\delta_{R}} \\ N_{\delta_{A}}^{*} + \frac{I_{XZ}}{I_{Z}} L_{\delta_{A}}^{*} & N_{\delta_{R}}^{*} + \frac{I_{XZ}}{I_{Z}} L_{\delta_{R}}^{*} \\ 0 & 0 \end{bmatrix}$ $U_c = \begin{vmatrix} \Delta \delta_A \\ \Delta \delta_A \end{vmatrix}$

> ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

18

STUDENTS-HUB.com

Linearization: Points to remember

- Linearized system is always a <u>local</u> <u>approximation</u> about the operating point
- As the operating point changes, the linearized model changes (for the same nonlinear system)
- The usual objective of control design using the linearized dynamics is "<u>deviation minimization</u>" (i.e. regulation)
- Control design based on linearized dynamics always relies on the philosophy of "gain scheduling" (i.e. gain interpolation)

ADVANCED CONTROL SYSTEM DESIGN Dr. Radhakant Padhi, AE Dept., IISc-Bangalore

STUDENTS-HUB.com

